

11. Use FFT to Design DF



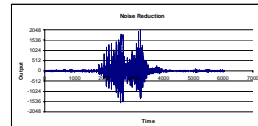
Lab 4 Application of RTOS

- Input sound, analog filter
- Digital filter, FFT
- Display amplitude versus freq on the oLED

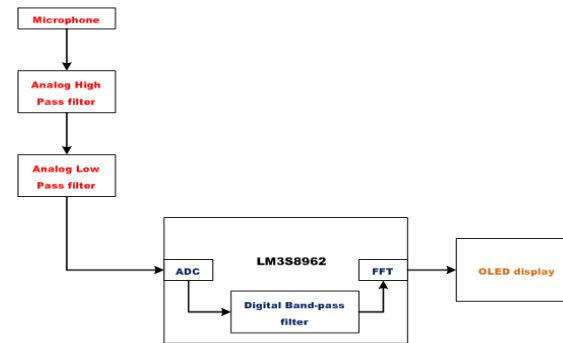
Reference EE345M book, chapter 5

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Lab4 Spectrum Analyzer

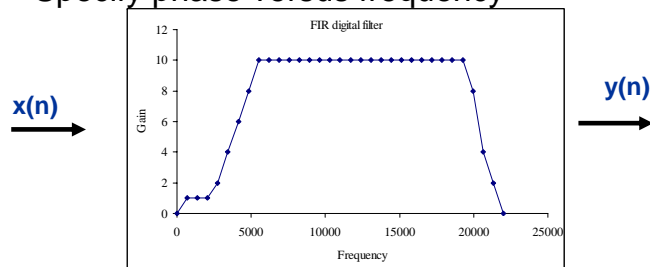


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Design a linear FIR digital filter

- Specify gain versus frequency
- Specify phase versus frequency

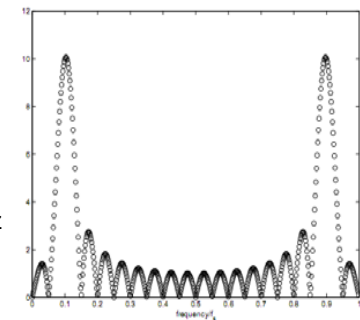


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Fast Fourier Transform (FFT)

- FFT is a faster version of the Discrete Fourier Transform (DFT)
- FFT spectrum of a cosine, which has a frequency of 0.1 F_s
 - $F_s = 10000$ Hz
 - Cosine freq = 1000 Hz
- Region interested in from 0 to $F_s/2$ (symmetric about $F_s/2$)

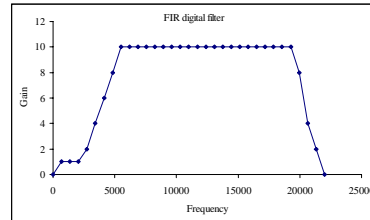


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FIR digital filter

- $Y(z) = H(z) X(z)$
- $h(n) = \text{IFFT} \{H(z)\}$
- Convolution
– $y(n) = h(n) * x(n)$



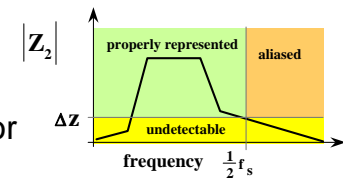
- Constants h_0, h_1, \dots, h_{N-1}
- $y(n) = h_0 \cdot x(n) + h_1 \cdot x(n-1) + \dots + h_{N-1} \cdot x(n-(N-1))$
- N multiplies, $N-1$ additions per sample

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How to choose sampling rate

- Nyquist Rate
- Limitation of display
- Limitation of processor
- Limitation of RAM
- Limitation of human eyes and ears
- Limitation of communication channel



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How to choose number of samples

- Frequency resolution = f_s/N
- Increase in N results in better frequency resolution
- However, increase in N leads to a bigger MACQ buffer and more multiplies and additions
- Does not need to be a power of 2
– DFT calculated once, off line

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Design process (1)

- Specify desired gain and phase, 0 to $\frac{1}{2} f_s$
 - k goes from 0 to $N/2$ ($f = k f_s/N$)
 - $H(k)$ is complex
 - $|H(k)|$ is gain
 - $\text{angle}(H(k))$ is phase
- For $\frac{1}{2} f_s$ to f_s
 - $H(N-k)$ is complex conjugate of $H(k)$
 - Poles and zeros are in complex conjugate pairs

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Design process (2)

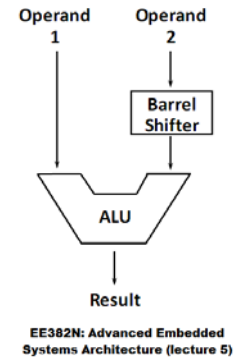
- Take IDFT of $H(k)$ to yield $h(n)$
 - n goes from 0 to $N-1$
 - $h(n)$ will be real, because $H(k)$ symmetric
- The digital filter is
 - $y(n) = h_0 \cdot x(n) + h_1 \cdot x(n-1) + \dots + h_{N-1} \cdot x(n-(N-1))$

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Binary Fixed point notation

- Binary fixed-point is faster than decimal fixed-point
- Qn Number (16 bit)
 - n : specifies the resolution = 2^{-n}
 - $16-n$: specifies Range
- Eg: 10.450 (unsigned number) with $n=11$?
- How is this number stored as an integer?



Value = Integer/2048

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10,450 \approx 01010.1110011010

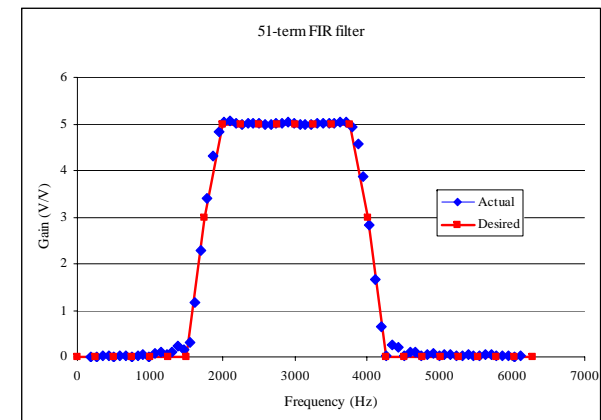
Example

- Open FIRdesign51.xls
- Change sampling rate to 10,000 Hz
- Adjust red desired gain to make BPF
 - Pass 2 to 4 kHz
 - Look at sharp corner versus round corner
- Notice linear phase
- Copy 51 coefficients into software

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2k to 4k BPF



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FIR Filter SW design

```

const long h[51]={-3,-9,4,5,0,17,5,-20,-5,-7,-22,
  24,41,-8,2,1,-74,-31,71,20,33,125,-119,-350,67,
  462,67,-350,-119,125,33,20,71,-31,-74,1,2,-8,41,
  24,-22,-7,-5,-20,5,17,0,5,4,-9,-3};
static unsigned int n=50; // 51,52,... 101
short Filter(short data){unsigned int k;
static long x[102]; // this MACQ needs twice
long y;
n++;
if(n==102) n=51;
x[n] = x[n-51] = data; // two copies of new data
y = 0;
for(k=0;k<51;k++){
  y = y + h[k]*x[n-k]; // convolution
}
y = y/256; // fixed point
return y;

```

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Circular Buffering

Array Index	Filter Coefficient Array h[]	Circular Buffer Array xcirc[]
0	$h[0]$	$x[n - \text{newest}]$
1	$h[1]$	$x[n - \text{newest} + 1]$
⋮	⋮	⋮
		$x[n - 1]$
<i>newest</i>		$x[n]$
<i>oldest</i>		$x[n - N + 1]$
		$x[n - N + 2]$
⋮	⋮	⋮
$N - 2$	$h[N - 2]$	$x[n - \text{newest} - 2]$
$N - 1$	$h[N - 1]$	$x[n - \text{newest} - 1]$

Source: "Communication system design using DSP algorithms" by Steven A. Tretter (chapter 3, page 73)

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Optimization

- Pointer implementations of MACQ faster
- Do not try and shift the data
- Convolution $x[n]*h[n]$ takes N multiplies, N-1 additions per sample
 - Can be optimized to N/2 multiplies
 - Coefficients are symmetric
- Assembly optimization with MLA
 - Multiply with accumulate

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