The University of Texas at Austin Dept. of Electrical and Computer Engineering Midterm #1

Date: March 11, 2020

Course: EE 445S Evans

Name:

Last,

First

- The exam is scheduled to last 75 minutes.
- Open books and open notes. You may refer to your homework assignments and the homework solution sets.
- Calculators are allowed. *Please disable all connections from your calculator to other electronic devices.*
- You may use any standalone computer system, i.e. one that is not connected to a network. *Please disable all wireless connections on your computer system(s).*
- Please turn off all cell phones.
- No headphones allowed.
- All work should be performed on the quiz itself. If more space is needed, then use the backs of the pages.
- **Fully justify your answers.** If you decide to quote text from a source, please give the quote, page number and source citation.

Problem	Point Value	Your score	Торіс
1	28		Filter Analysis
2	24		Sampling
3	24		Audio Filter Design
4	24		Potpourri
Total	100		

Problem 1.1 Filter Analysis. 28 points.

Consider the following causal linear time-invariant (LTI) discrete-time filter with input x[n] and output y[n] described by

$$y[n] = a x[n] + b x[n-1] + c x[n-2]$$

for $n \ge 0$. Coefficients *a*, *b* and *c* are real-valued. In addition, $a \ne 0$ and $c \ne 0$.

(a) Is this a finite impulse response (FIR) or infinite impulse response (IIR) filter? Why? 3 points.

(b) What are the initial conditions and their values? Why? 6 points.

(c) Draw the block diagram of the filter relating input x[n] and output y[n]. 6 points.

(d) Derive a formula for the transfer function in the z-domain and the region of convergence. 4 points.

- (e) Give a formula for the discrete-time frequency response of the filter. Justify the steps. 3 points.
- (f) Determine formulas for the relationships among the filter coefficients to make the filter have generalized linear phase over all frequencies. Give a numeric value for each coefficient to achieve generalized linear phase, and indicate what the frequency selectivity is. *Hint*: Generalized linear phase means that the impulse response is odd symmetric about its midpoint. *6 points*.

Problem 1.2. Sampling. 24 points.

For each problem below, determine the frequency (or frequencies) present in x(t) and y(t) as well as the single sampling rate you would use for the entire system to prevent aliasing.

Please note that $T_c = 1 / f_c$ and $T_0 = 1 / f_0$ in the following. Each problem is worth 6 points.

(a) Let $x(t) = \cos(2\pi f_c t)$ be a continuous-time signal for $-\infty < t < \infty$.

$$\xrightarrow{x(t)} (\bullet)^2 \xrightarrow{y(t)}$$

(b) Let $x(t) = \cos(2\pi f_c t)$ be a continuous-time signal for $-\infty < t < \infty$.

$$\xrightarrow{x(t)} \underbrace{(\bullet)^4} \xrightarrow{y(t)}$$

(c) Let

 $x(t) = \operatorname{sinc}\left(\frac{t}{T_0}\right)$

be a continuous-time signal for $-\infty < t < \infty$ whose continuous-time Fourier transform is

$$X(f) = T_0 \operatorname{rect}\left(\frac{f}{f_0}\right)$$

Here, $f_c > f_0$

$$x(t) \xrightarrow{y(t)} y(t)$$

$$cos(2\pi f_c t)$$

(d) Let

$$x(t) = \cos(2\pi f_c t) \operatorname{sinc}\left(\frac{t}{T_0}\right)$$

be a continuous-time signal for $-\infty < t < \infty$
where $f_c > f_0$
$$\xrightarrow{x(t)} \underbrace{y(t)}_{cos(2\pi f_c t)}$$

Problem 1.3 Audio Filter Design. 24 points.

This problem asks you to evaluate tradeoffs in two designs for a filter for a tweeter/treble speaker:

- Speaker plays frequencies from roughly 2,000 Hz to 20,000 Hz.
- A discrete-time highpass filter will be placed in the speaker before the digital-to-analog (D/A) converter
- The D/A converter operates at a sampling rate of 48,000 Hz.

Highpass filter design specifications:

- Stopband frequency of 1800 Hz and passband frequency of 2000 Hz
- Stopband attenuation of 80 dB and passband tolerance of 1 dB
- Sampling rate of 48,000 Hz.

Proposed filter design #1: Finite Impulse Response (FIR) Filter.

Parks-McClellan (Equiripple) design. Order = 660. Meets specifications.

Proposed filter design #2: Infinite Impulse Response (IIR) Filter.

Elliptic (Equiripple) design. Order = 10. Meets specifications.



- (a) Assuming the FIR filter is in direct form and the IIR filter is in a cascade of biquads (second-order sections), compute the number of multiplications per sample required by each. *6 points*.
- (b) For the IIR filter design, the group delay for frequencies greater than 4,000 Hz is less than 7 samples. What is the group delay for the FIR filter in the same range? *6 points*
- (c) For the IIR filter design, the largest group delay of 64–500 samples occurred over the range of 2000 Hz to 2200 Hz. Is there a way you would recommend to alter the filter specifications so that the group delay would be less than 64 throughout the entire passband? *6 points*
- (d) Which proposed filter design would you advocate using? 6 points.

Problem 1.4. Potpourri. 24 points.

- (a) You'd like to design a low-complexity lowpass finite impulse response (FIR) filter with an integer group delay. The two-tap averaging filter is a low-complexity lowpass FIR filter, but it has a group delay of ½ sample. Design two different low-complexity lowpass FIR filters with integer group delays based on the two-tap averaging filter. 6 points.
- (b) Your system only has the ability to generate half of the carrier frequency you need for a communication system. What signal processing operations would you add to generate the carrier frequency? Draw a block diagram for your approach. *6 points*.
- (c) We can use partial fractions decomposition to convert a transfer function into a parallel implementation. Consider a second-order system with conjugate symmetric poles p_0 and p_1 and conjugate symmetric zeros z_0 and z_1 . We can rewrite the second-order system as a sum of two first-order sections assuming that the poles are not equal:

$$H(z) = \frac{(1 - z_0 z^{-1})(1 - z_1 z^{-1})}{(1 - p_0 z^{-1})(1 - p_1 z^{-1})} = \frac{1 - c_0 z^{-1}}{1 - p_0 z^{-1}} + \frac{1 - c_1 z^{-1}}{1 - p_1 z^{-1}}$$

Please note that constants c_0 and c_1 are complex-valued.

- i. How many real-valued multiplications per output sample are needed for the second-order system? *3 points*.
- ii. How many real-valued multiplication operations per output sample are needed for the parallel combination of the two first-order sections? *3 points*.
- iii. Assuming that the two first-order sections can be executed in parallel, which realization requires fewer real-valued multiplications per output sample to compute? *3 points*.
- iv. Repeat part iii assuming that poles p_0 and p_1 , zeros z_0 and z_1 , and constants c_0 and c_1 are real-valued. *3 points*.