The University of Texas at Austin Dept. of Electrical and Computer Engineering Midterm #2

Prof. Brian L. Evans

Date: May 4, 2018

Course: EE 445S

Name:

Last,

First

- The exam is scheduled to last 50 minutes.
- Open books and open notes. You may refer to your homework assignments and the homework solution sets. You may not share materials with other students.
- Calculators are allowed.
- You may use any standalone computer system, i.e. one that is not connected to a network. **Disable all wireless access from your standalone computer system**.
- Please turn off all smart phones and other personal communication devices.
- Please remove headphones.
- All work should be performed on the quiz itself. If more space is needed, then use the backs of the pages.
- Fully justify your answers unless instructed otherwise. When justifying your answers, you may refer to the Johnson, Sethares & Klein (JSK) textbook, the Welch, Wright and Morrow (WWM) lab book, course reader, and course handouts. Please be sure to reference the page/slide number and quote the particular content in your justification.

Problem	Point Value	Your score	Торіс
1	24		Baseband Pulse Amplitude Modulation
2	28		QAM Communication Performance
3	24		QAM Constellation Derotation
4	24		Potpourri
Total	100		

Problem 2.1. Baseband Pulse Amplitude Modulation. 24 points.

A baseband pulse amplitude modulation transmitter is described as



For $N_g = 4$ and L = 20, a plot is shown below for 10 symbol periods over 0 to 30 ms of s[m] after it had passed through a digital-to-analog converter, and the symbol amplitudes a[n] are shown as a stem plot:



(a) What is the value of J, the number of bits per symbol? Why? 3 points.

- (b) If the spacing between constellation points is 2d, what is the value of d? Why? 3 points.
- (c) Draw a constellation map with Gray coding. 3 points.
- (d) Accurately compute the symbol time, T_{sym} , in milliseconds. 3 points.
- (e) Give a formula for the pulse shape, g[m]. How many samples are in g[m]? 6 points.
- (f) Infer an upper bound on the amplitude of s[m] as a function of d, J and N_g . 6 points.

Problem 2.2 QAM Communication Performance. 28 points.

Consider the two 16-QAM constellations below. Constellation spacing is 2d.



Energy in the pulse shape is 1. Symbol time T_{sym} is 1s. The constellation on the left includes the decision regions with boundaries shown by the in-phase (I) axis, quadrature (Q) axis and dashed lines.

Euch pur below is worth 5 points. I lease fully justify your unswer	Each p	art below i	is worth 3	points.	Please fully	y justify	your answer
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	Left Constellation	Right Constellation				
(a) Peak transmit power	$18d^{2}$					
(b) Average transmit power	$10d^{2}$					
(c) Draw the decision regions for the right constellation on top of the right constellation.						
(d) Number of type I regions	4					
(e) Number of type II regions	8					
(f) Number of type III regions	4					
(g) Probability of symbol error	$3O(d) - {}^{9}O^{2}(d)$					
for additive Gaussian noise	$SQ(\overline{\sigma})^{-}\overline{4}Q(\overline{\sigma})$					
with zero mean & variance σ^2						

(h) Consider using the constellations in upconverted QAM. In the QAM receiver, how would the Costas loop for the phase locked loop perform for the right constellation vs. the left constellation? *7 points*.

Problem 2.3. QAM Constellation Derotation. 24 points.

A baseband Quadrature Amplitude Modulation (QAM) receiver is given below



where $\hat{i}[n]$ and $\hat{q}[n]$ are the received in-phase and quadrature symbol amplitudes at symbol index *n*. At the receiver, the QAM constellation may rotate due to a mismatch in the carrier frequencies.

A phase locked loop running at the sampling rate could track the time-varying phase that is due to the carrier frequency mismatch.

An alternative is to derotate the constellation at the symbol rate by multiplying the complex symbol $\hat{i}[n] + j \hat{q}[n]$ by $e^{j\theta}$, i.e. $i[n] + j q[n] = (\hat{i}[n] + j \hat{q}[n])e^{j\theta} = (\hat{i}[n] + j \hat{q}[n])(\cos \theta + j \sin \theta)$:

 $i[n] = \hat{\imath}[n] \cos \theta - \hat{q}[n] \sin \theta$ and $q[n] = \hat{q}[n] \cos \theta + \hat{\imath}[n] \sin \theta$

We will adapt the phase offset θ based on the error vector magnitude e[n] in the decision device, i.e.

$$e^{2}[n] = (i[n] - \hat{i}[n])^{2} + (q[n] - \hat{q}[n])^{2}$$

(a) Give an objective function J(e[n]). 6 points.

- (b) Derive the update equation for θ_{k+1} , where k is a symbol index. 9 points.
- (c) What range of values would you recommend for μ ? 3 points.
- (d) This method can work with or without a training sequence. If you were to use a training sequence, which one would you use? Why? 6 points.

Problem 2.4. Potpourri. 24 points

(a) What is the primary advantage of using symbol amplitudes of -3*d*, -*d*, *d* and 3*d* for 4-level pulse amplitude modulation instead of *d*, 3*d*, 5*d*, and 7*d*? 6 points.

(b) How will fifth-generation (5G) cellular communication systems be able to provide 10 times the average and peak bit rates of fourth-generation (4G) cellular communication systems? *6 points*.

- (c) For each communication subsystem below, advocate using either a discrete-time digital implementation or a continuous-time analog implementation.
 - *i.* Baseband processing. *3 points.*
 - *ii.* Upconversion to carrier frequencies greater than 1 GHz. *3 points*.
- (d) In the automatic gain control (AGC) block diagram given below, the analog-to-digital converter outputs r[m] which is a signed integer of *B* bits. Give a formula that uses r[m] and the adapted gain c(t) to create a floating-point approximation of $r_1(t)$. This type of floating-point analog-to-digital conversion is used in practice, e.g. in cellular basestations. *6 points*.

