

2000 IEEE Digital Signal Processing Workshop

Fast Time-Domain Equalization for Discrete Multitone Modulation Systems

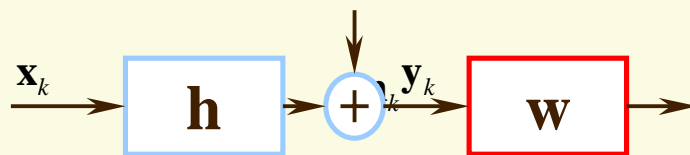
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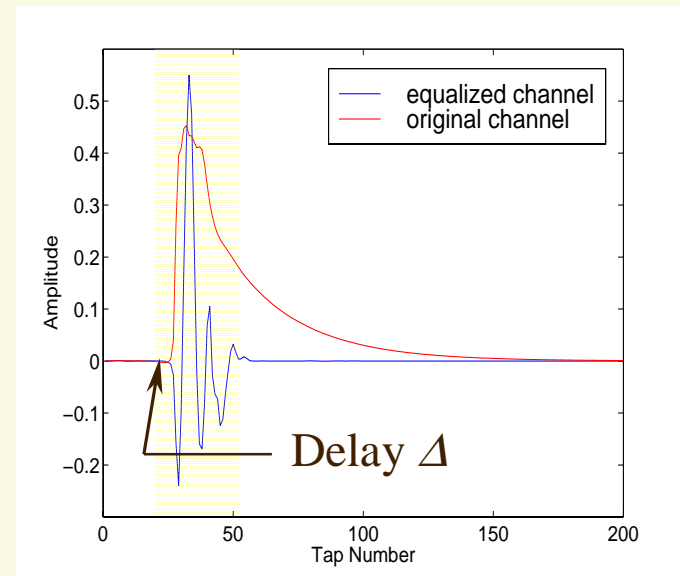
Problem Statement

✓ Effective channel impulse response



\mathbf{h} : L_h -tap channel impulse response

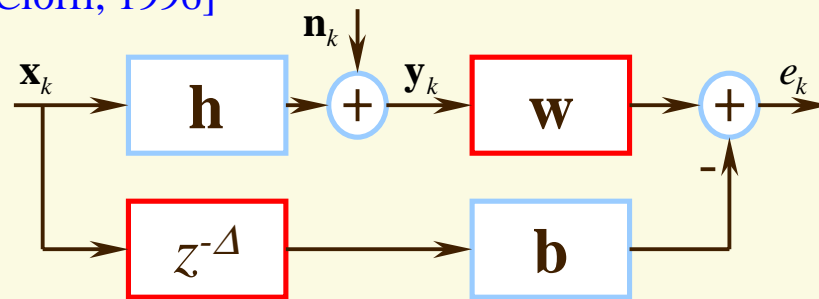
\mathbf{w} : N_w -tap time-domain equalizer (TEQ)



- ✓ Problem: High computational cost of optimal TEQ design during ADSL transceiver initialization
- ✓ Goal: Low complexity suboptimal TEQ design

Minimum Mean Squared Error Method

- ✓ Minimize MSE [Falconer & Magee, 1973][Chow & Cioffi, 1992][Al-Dhahir & Cioffi, 1996]



$$\mathbf{w} = [w(1) \quad \cdots \quad w(N_w)]^T$$

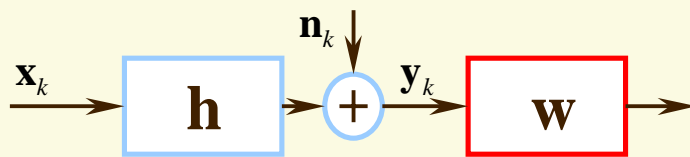
$$\mathbf{b} = [b(0) \quad b(1) \quad \cdots \quad b(\nu)]^T$$

$$\text{MSE} = E \left\{ (\mathbf{w}^T \mathbf{y}_k - \mathbf{b}^T \mathbf{x}_{k-\Delta})^2 \right\}$$

- ✓ Constraints to avoid trivial solution
 - Unit tap constraint: $b(i) = 1, i \in \{0, 1, \dots, \nu\}$
 - Unit norm constraint: $\|\mathbf{b}\| = 1$ or $\|\mathbf{w}\| = 1$
- ✓ Computational cost for each Δ
 - Matrix inversion
 - Eigenvalue decomposition

Maximum Shortening SNR Method

- ✓ Minimize energy outside a window of length $(\nu+1)$ [Melsa, Younce, & Rohrs, 1996]

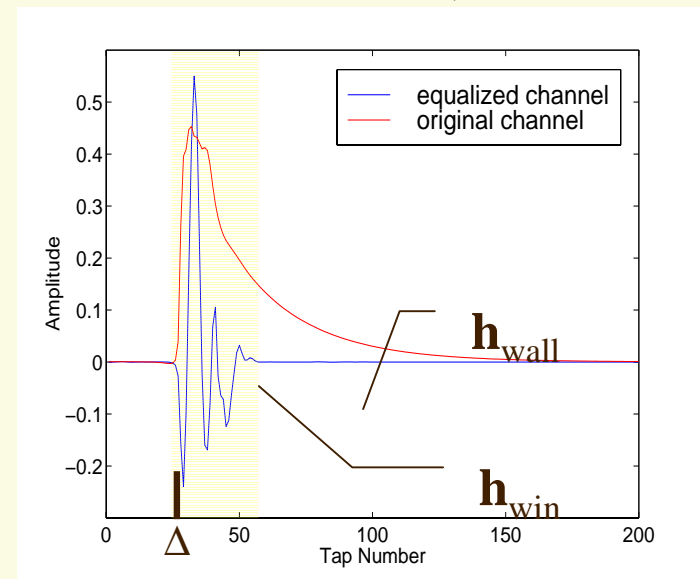


$$\mathbf{h}_{\text{wall}}^T \mathbf{h}_{\text{wall}} = \mathbf{w}^T \mathbf{H}_{\text{wall}}^T \mathbf{H}_{\text{wall}} \mathbf{w} = \mathbf{w}^T \mathbf{A} \mathbf{w}$$

$$\mathbf{h}_{\text{win}}^T \mathbf{h}_{\text{win}} = \mathbf{w}^T \mathbf{H}_{\text{win}}^T \mathbf{H}_{\text{win}} \mathbf{w} = \mathbf{w}^T \mathbf{B} \mathbf{w}$$

$$\text{SSNR} = 10 \log_{10} \left(\frac{\mathbf{w}^T \mathbf{B} \mathbf{w}}{\mathbf{w}^T \mathbf{A} \mathbf{w}} \right)$$

- ✓ Constraint: $\mathbf{w}^T \mathbf{B} \mathbf{w} = 1$
- ✓ Computational cost for each Δ
 - Matrix inversion
 - Cholesky decomposition
 - Eigenvalue decomposition



Motivation

✓ MMSE method

- Minimizes MSE both inside window and outside window
- Slow Convergence

✓ Maximum SSNR method

- Requires high computational cost

✓ Both methods search for optimal delay Δ

$$0 \leq \Delta \leq L_h + N_w - \nu - 2 \Rightarrow 0 \leq \Delta \leq 512 + 21 - 32 - 2 = 499$$

Divide-and-Conquer TEQ

- ✓ Use Divide-and-Conquer algorithm to minimize $\mathbf{h}_{\text{wall}}^T \mathbf{h}_{\text{wall}}$
- ✓ Divide a N_w -tap TEQ filter into (N_w-1) two-tap TEQs
- ✓ Each two-tap filter is initialized as

$$\mathbf{w}_i = [1, g_i]^T$$

- ✓ Calculate each g_i at i^{th} iteration

$$g_i = -\frac{\sum_{k \in S} \tilde{h}_{i-1}(k-1)\tilde{h}_{i-1}(k)}{\sum_{k \in S} \tilde{h}_{i-1}^2(k-1)}, S = \{1, 2, \dots, \Delta, \Delta + \nu + 2, \dots, L_{\tilde{h}_{i-1}}\}$$

- ✓ Convolve all two-tap filters to obtain the N_w -tap TEQ

Heuristic Search of Delay Δ

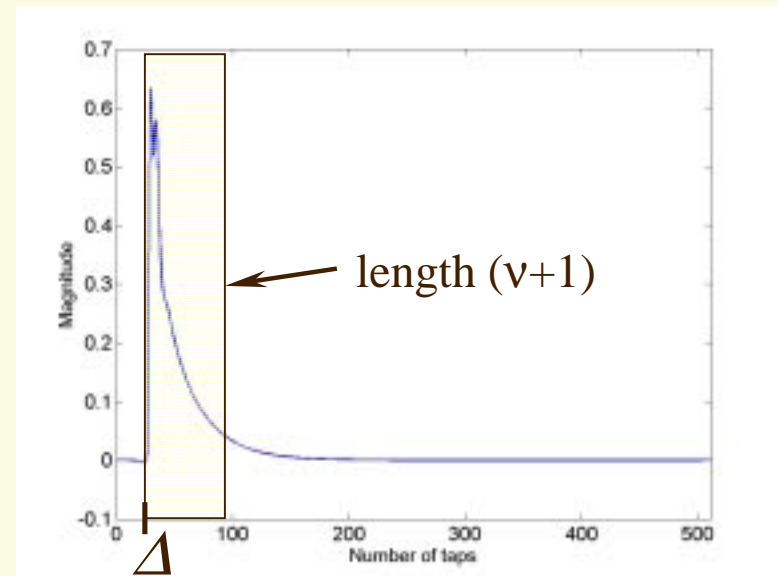
- ✓ Estimate the optimal delay

$$\Delta_{\text{ratio}} = \arg \max_{\Delta} \frac{\text{energy inside a window of original } \mathbf{h}}{\text{energy outside a window of original } \mathbf{h}}$$

- ✓ Computational cost

- L_h multiplications
- $(L_h - 2)\Delta_{\text{range}}$ additions
- Δ_{range} division

Δ_{range} : the number of Δ s to be searched



Computer Simulation

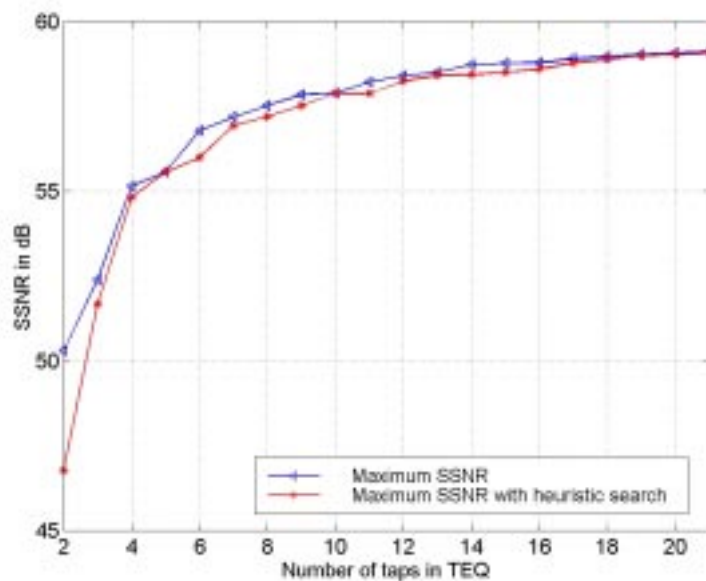
- ✓ Channel: carrier-serving-area digital subscriber loop 1
- ✓ Sampling rate: 2.208 MHz
- ✓ Cyclic prefix: $\nu = 32$
- ✓ Number of samples per symbol: 512+32
- ✓ Equalizer taps: $N_w = 21$

Methods	\times	$+$	\div
Maximum SSNR	120379	118552	441
Divide-and-Conquer TEQ	41000	40880	20

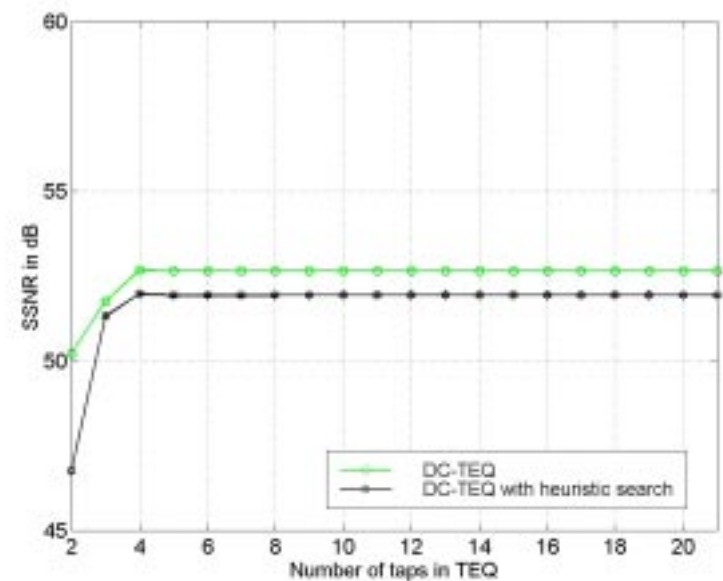
Computations for each value of delay Δ considered

Simulation Results

Maximum SSNR



DC-TEQ



- ✓ Heuristic search: 1 dB loss for four or more TEQ taps
- ✓ Divide-and-conquer TEQ: 4-8 dB loss of SSNR

Conclusion

✓ Derive Divide-and-Conquer TEQ design method

- Requires fewer computations than maximum SSNR method
- For G.DMT ADSL, $L_h = 512$, $\nu = 32$, and $N_w = 21$,
 - » Reduces multiplications and additions by a factor of 3
 - » Reduces divisions by a factor of 22
- Comparable performance to the maximum SSNR method

✓ Develop heuristic search to find delay

- Find Δ by maximizing energy ratio of original channel impulse response
- Reduces computational complexity in return for minimal performance loss
 - » Loss is 4 dB for 2-tap TEQ and 1 dB for 4-tap TEQ