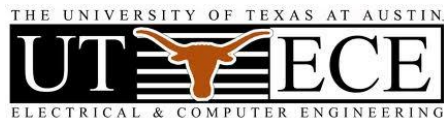


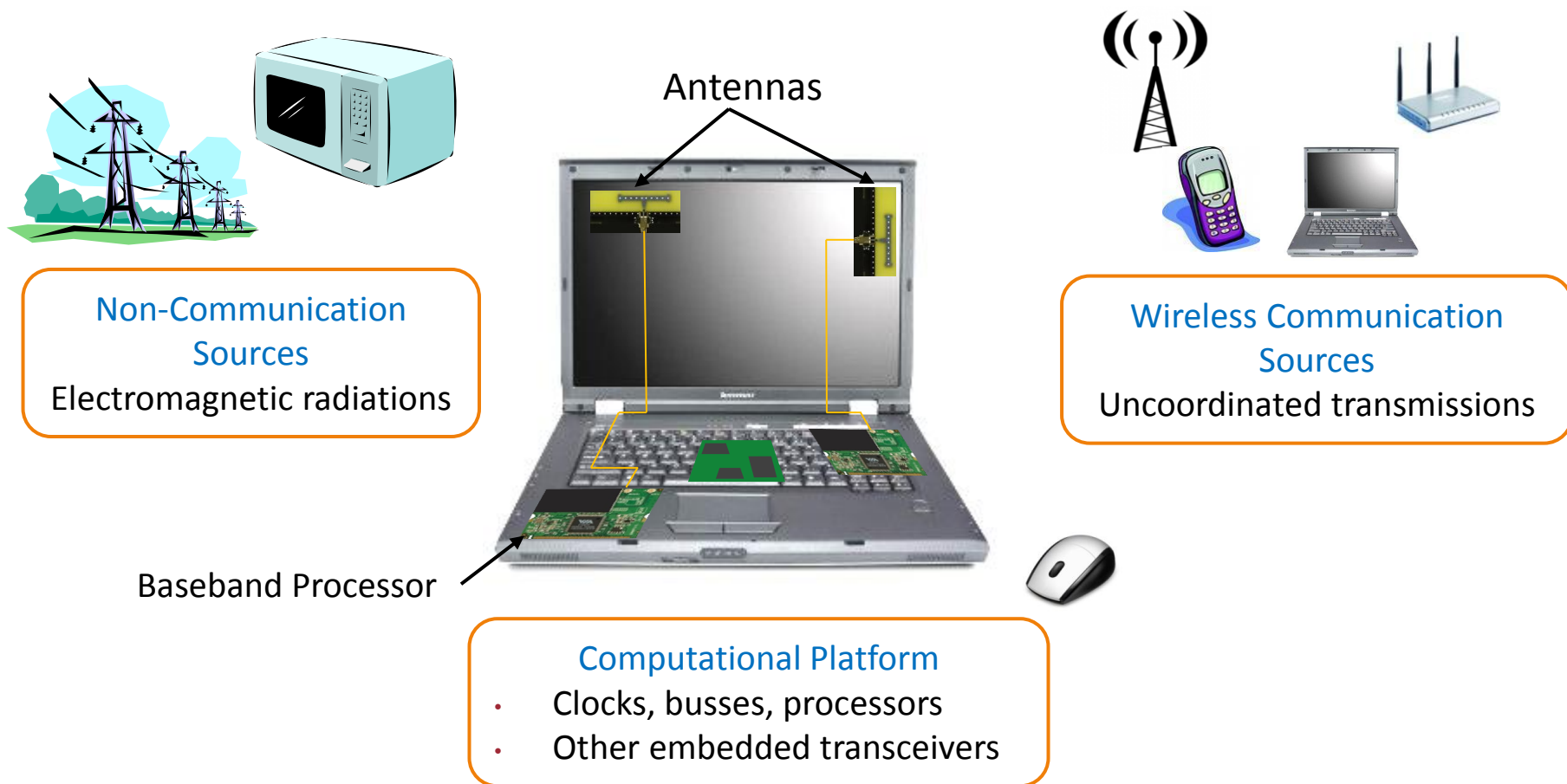
IMPULSIVE NOISE MITIGATION IN OFDM SYSTEMS USING SPARSE BAYESIAN LEARNING

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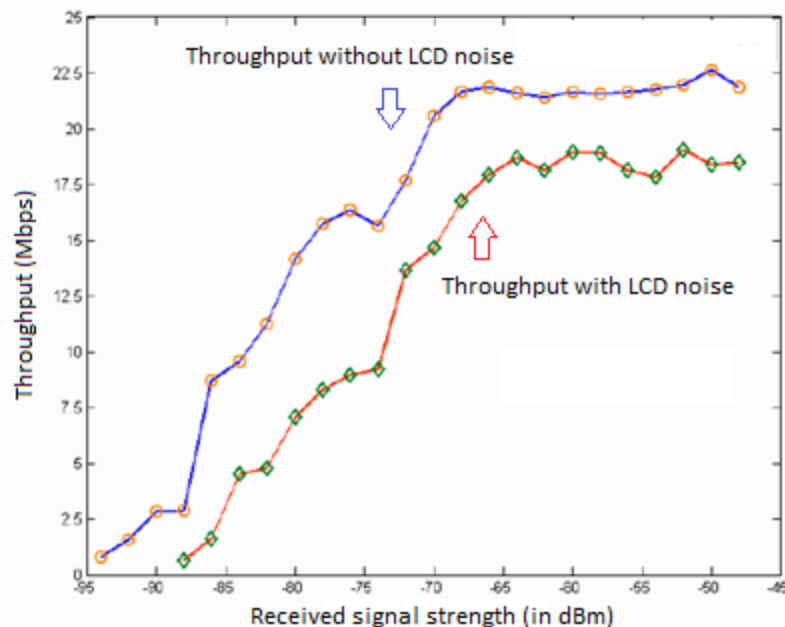


Impulsive Noise at Wireless Receivers

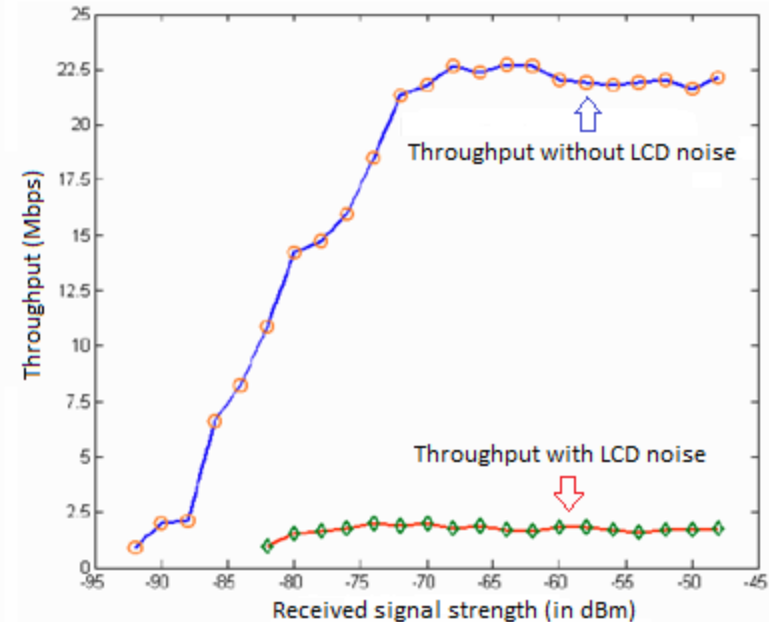


In-Platform Interference

- May severely degrade communication performance
- Impact of LCD noise on throughput for IEEE 802.11g embedded wireless receiver ([Shi2006])

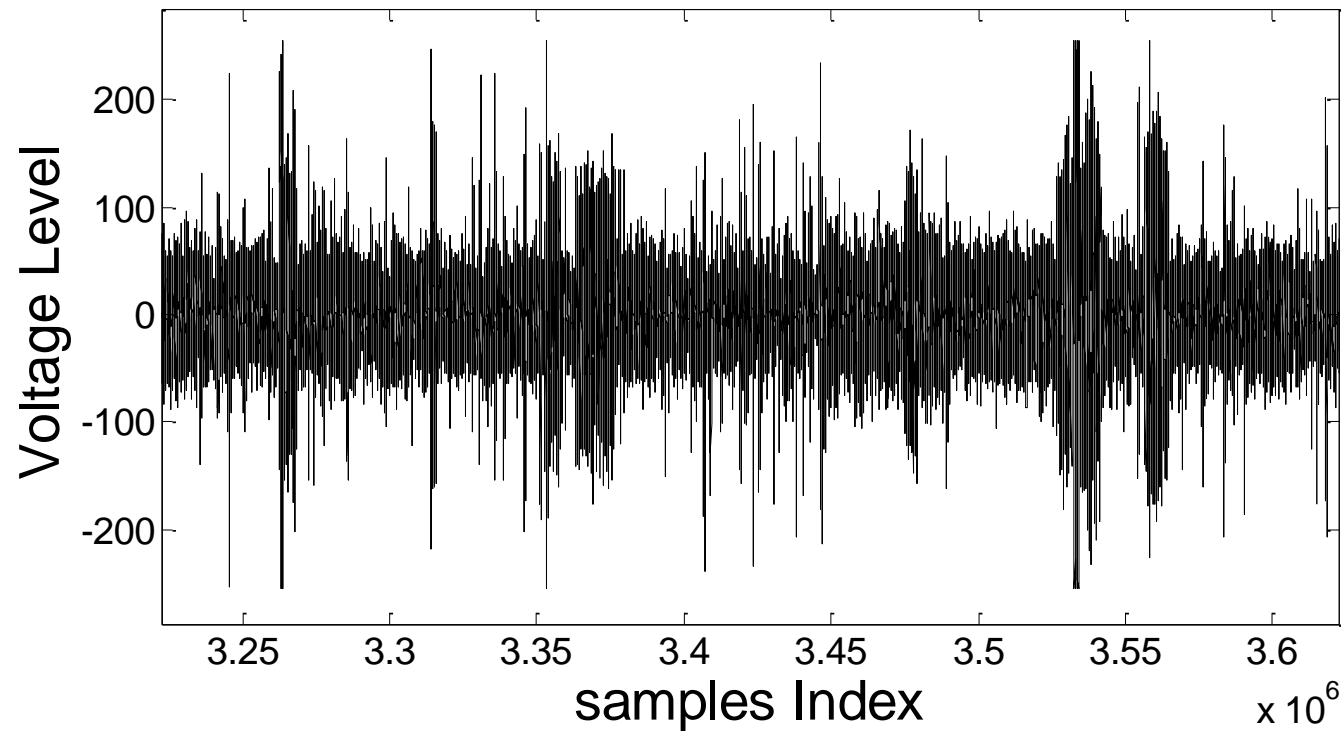


(a) 802.11g, Channel 1

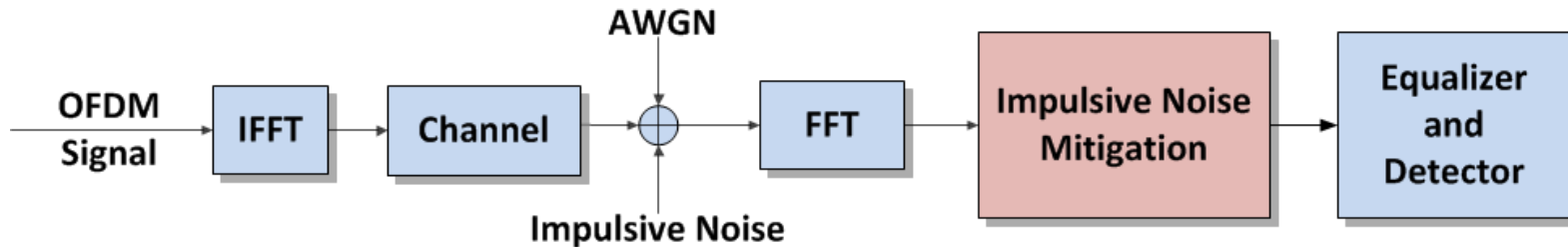


(b) 802.11g, Channel 7

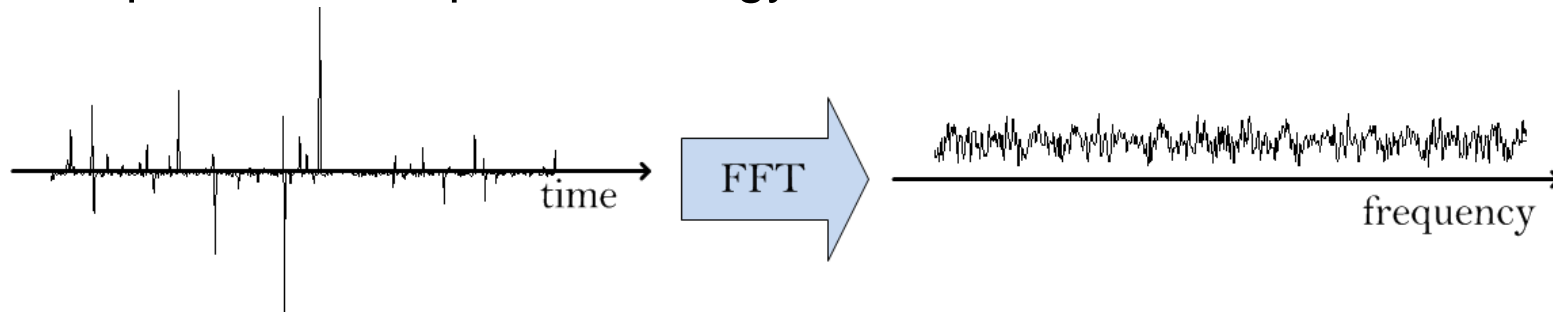
Noise Trace for Platform Noise



OFDM System in Impulsive Noise



- FFT spreads out impulsive energy across all tones



- SNR of each tone is decreased
- Receiver performance degrades
- Noise in each tone is asymptotically Gaussian (as $N_{DFT} \rightarrow \infty$)

Prior Work

- Parametric vs. non-parametric methods (Noise Statistics)

	Param.	Non-Param.
Assume parameterized noise statistics	Yes	No
Performance degradation due to model mismatch	Yes	No
Training needed	Yes	No

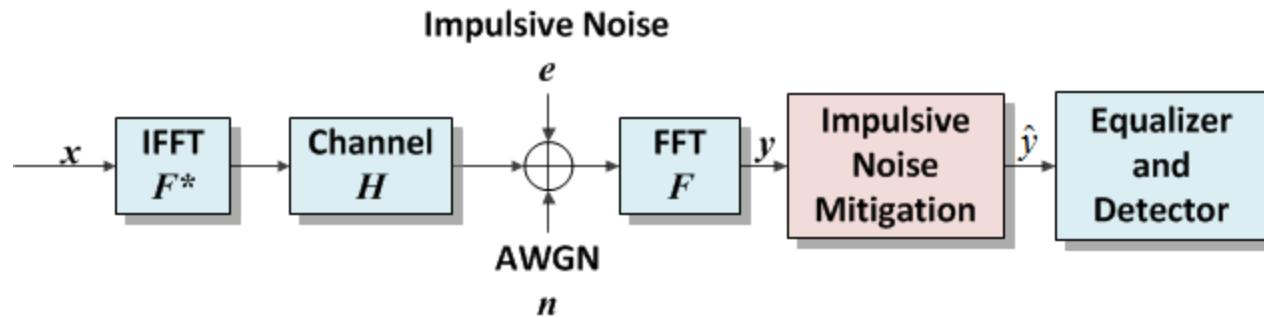
- Impulsive noise mitigation in OFDM

	Parametric?	Technique	Optimality	Complexity	
	[Nassar09]	Yes	Pre-filtering	★☆☆☆	★☆☆☆
✓	[Haring02]	Yes	MMSE** estimate	★★★☆☆	★★★☆☆
	[Haring03]	Yes	Iterative decoder	★★★★☆	★★★★☆
✓	[Caire08]	No*	Compressed sensing & LS**	★★☆☆☆	★★★☆☆

* Semi-non-parametric since threshold tuning is needed

** MMSE: Minimum mean squared error; LS: least squares

System Model



- A linear system with Gaussian disturbance

$$y = Fe + \underbrace{FHF^*x}_{\Lambda} + \underbrace{Fn}_g = Fe + v, \quad v \sim CN(\Lambda x, \sigma^2 I)$$

- ✓ Estimate the impulsive noise and remove it from the received signal

$$\hat{y} = y - F\hat{e} = \Lambda x + g + F(\hat{e} - e) \xrightarrow{\hat{e} \approx e} \Lambda x + g$$

- ✓ Apply standard OFDM decoder as if only Gaussian noise were present
- ✓ Goal: Non-parametric impulsive noise estimator

Estimation Using Null Tones

- Underdetermined linear regression

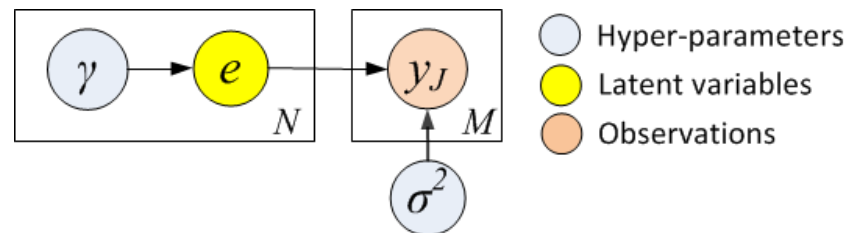
- F_J : over-complete dictionary
- e : sparse weight vector
- $g \sim CN(0, \sigma^2 I)$ with σ^2 unknown

$$y_J = F_J e + g$$

J : Index set of null tones
($x_j = 0$)

- Sparse Bayesian learning (SBL)

- Prior: $e | \gamma \sim CN(0, \Gamma), \Gamma \triangleq \text{diag}(\gamma)$
- Likelihood: $y_J | \gamma, \sigma^2 \sim CN(0, F_J F_J^* + \sigma^2 I)$
- Posterior: $e | y_J; \gamma, \sigma^2 \sim CN(\mu, \Sigma_e)$



Step 1: Maximum likelihood estimate of hyper-parameters: $(\hat{\gamma}, \hat{\sigma}^2) = \underset{\gamma, \sigma^2}{\text{argmax}} p(y_J; \gamma, \sigma^2)$

Treat e as latent variables and solve by expectation maximization (EM)

$\hat{\gamma}$ and $\hat{\sigma}^2$ are inter-dependent and updated iteratively

Step 2: Estimate e from posterior mean: $\hat{e} = E[e | y_J; \hat{\gamma}, \hat{\sigma}^2] = \hat{\mu}$

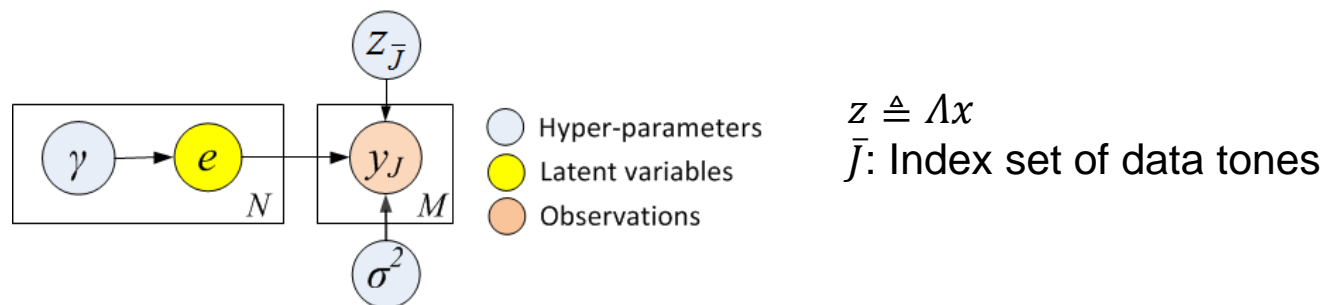
Guaranteed to converge to a sparse solution.

Estimation Using All Tones

- Joint estimation of data and noise

$$y = Fe + v \quad v \sim CN(\Lambda x, \sigma^2 I)$$

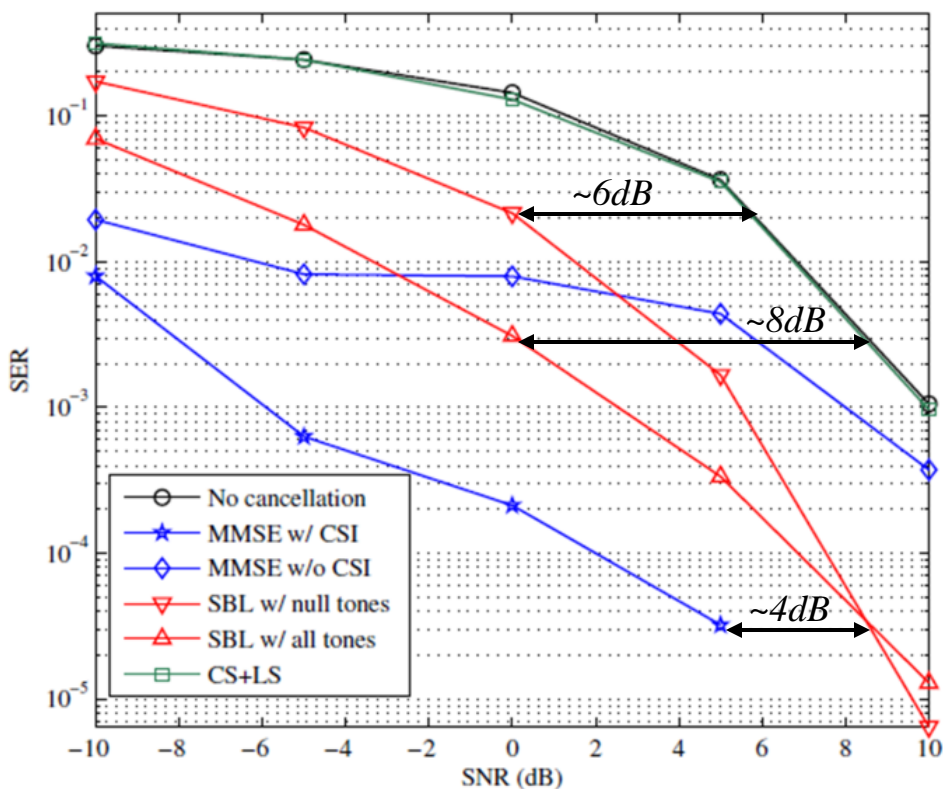
- Similar SBL approach with additional hyper-parameters of the data



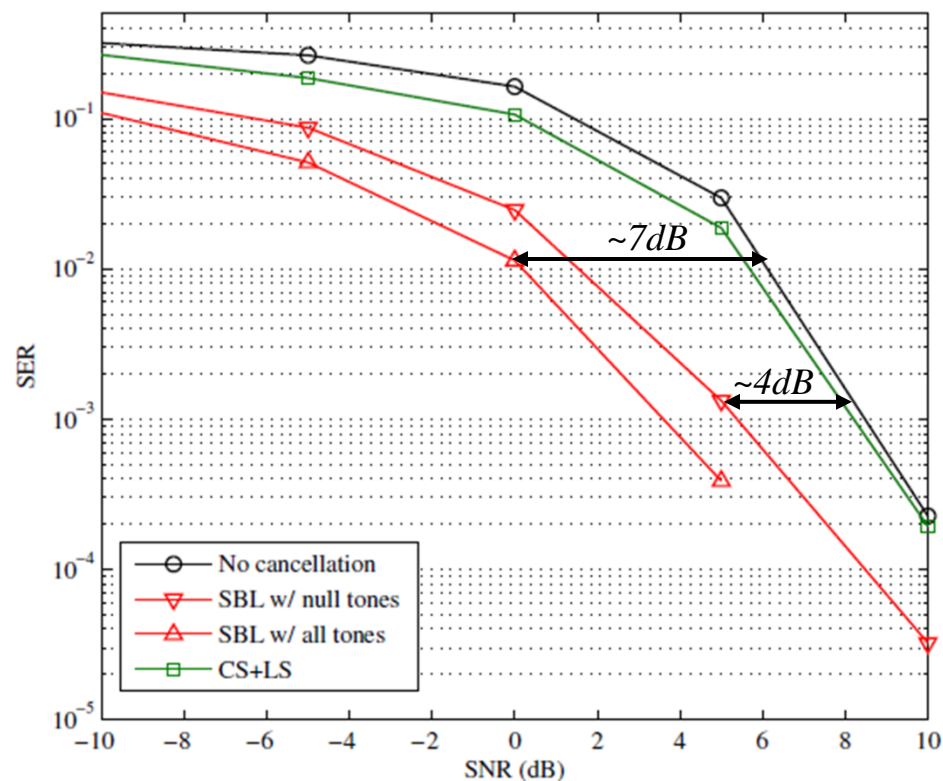
- Step 1 involves ML optimization over 3 sets of hyper-parameters: $(\gamma, \sigma^2, (\Lambda x)_{\bar{J}})$
- $x_{\bar{J}}$ is relaxed to be continuous variables to insure a tractable M-step
- Estimate of $(\Lambda x)_{\bar{J}}$ is sent to standard OFDM channel equalizer and MAP detector
- Increase complexity from $O(N^2M)$ to $O(N^3)$ per EM iteration

Simulation Results

- Symbol error rate (SER) performance in different noise scenarios



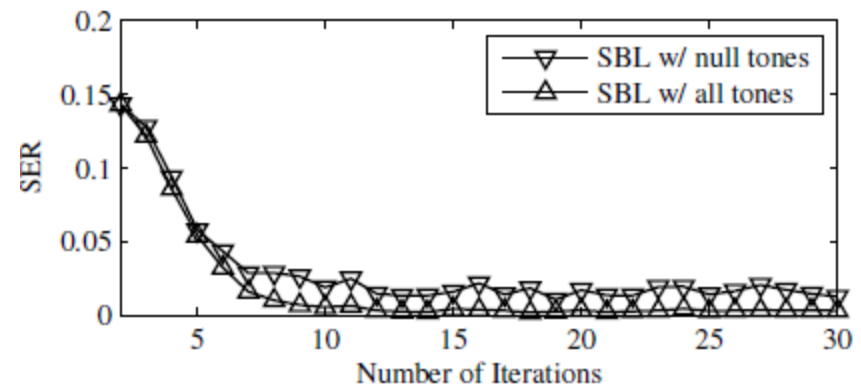
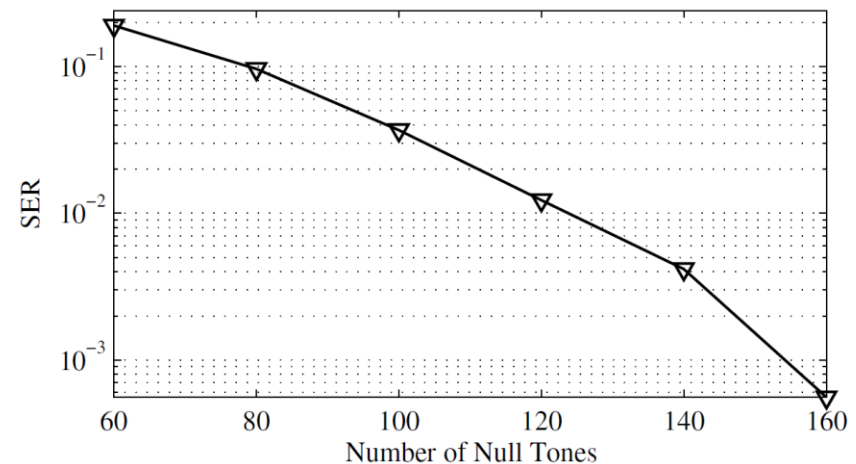
Middleton's Class A model



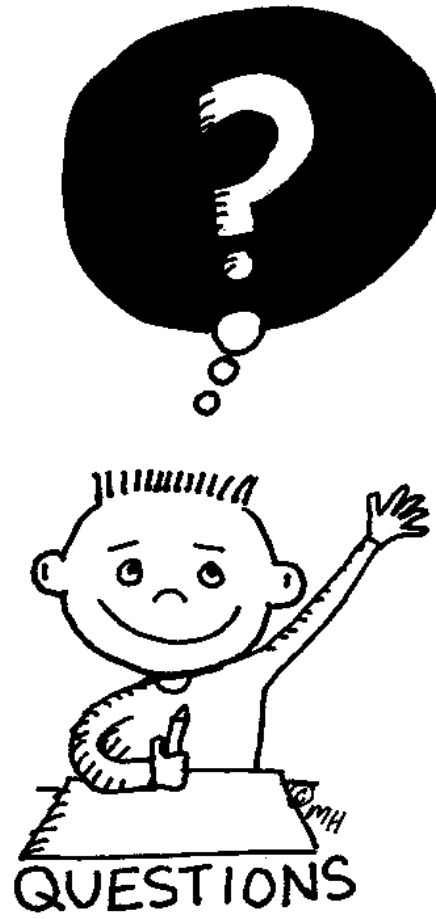
Symmetric alpha stable model

Simulation Results

- Performance of the first algorithm vs. the number of null tones
 - SNR = 0dB
 - 256 tones
 - Middleton Class A noise
- In both algorithms, EM converges after a few iterations



Thank you for your attention!



References

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