# **Multi-Criteria Analog IIR Filter Optimization**



#### **Prof. Brian L. Evans**

http://www.ece.utexas.edu/~bevans



work performed with Niranjan Damera-Venkata

## Electrical and Computer Engineering Department The University of Texas at Austin Austin, TX 78712-1084

http://anchovy.ece.utexas.edu

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### Outline

- Introduction
- Modeling
- Objective Measures for Properties
- Distance Measures for Properties
- Design Example
- Validation
- Conclusion

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#### Introduction

- *Problem*: Optimize multiple analog filter behavioral and implementation characteristics at the same time
- Goal: Develop an extensible, automated framework
- Solution: Filter Optimization Packages for Mathematica
  - Constrained non-linear optimization as Sequential Quadratic Programming: converges to global optimum & robust when closed-form gradients provided.
  - Program Mathematica to derive formulas for cost function, constraints, and gradients, and convert the formulas to Matlab programs to run optimization.
  - Example: linearize phase and minimize peak overshoot of an elliptic filter; constraining  $Q_{max}$  to 10 reduced  $Q_{max}$  from 61 to 10 (filter easier to build)



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#### Modeling

- Free Parameters: Locations of Poles and Zeros
  - List of *n* conjugate pole-pairs  $a_k \pm j b_k$  (and multiplicities)
  - List of *r* conjugate zero-pairs  $c_k \pm j d_k$  (and multiplicities)
  - Can be combined into a cascade of second-order sections

#### Properties

- Behavioral: magnitude, phase, and step responses
- Implementation: quality factors
- All properties are real-valued

### • Formulate Optimization Problem

- *Objective measures* of the properties as functions of the free parameters
- Distance measures for deviation of the actual and desired property values
- Cost function as a weighted combination of distance measures
- Constraints on the values of the free parameters

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#### **Objective Measures for Properties**

**Objective Measures for the All-Pole IIR Filter Case** 

• Magnitude response (with polynomials in Horner's form)

$$|H(j\omega)| = \prod_{k=1}^{n} \frac{a_k^2 + b_k^2}{\sqrt{(\omega^2 + 2(a_k^2 - b_k^2))\omega^2 + (a_k^2 + b_k^2)^2}}$$

Unwrapped phase response

$$\angle H(j\omega) = \sum_{k=1}^{n} \left( \operatorname{atan} \frac{\omega - b_k}{a_k} + \operatorname{atan} \frac{\omega + b_k}{a_k} \right)$$

- Quality factors
  - For kth second-order section, use standard formula
  - $Q_k >= 0.5$ , where  $Q_k = 0.5$  corresponds to a real pole and  $Q_k$  of infinity corresponds to an imaginary pole

$$Q_k = \frac{\sqrt{a_k^2 + b_k^2}}{-2a_k}$$

• Effective quality factor  $Q_{eff}$  is a combination of the second-order quality factors: we chose the *geometric mean* 

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**Measuring Peak Overshoot in the Step Response** 

Partial fractions decomposition

$$\frac{H(s)}{s} = \frac{1}{s} \sum_{k=1}^{n} \frac{C_k s + D_k}{s^2 - 2a_k s + a_k^2 + b_k^2}$$

• Step response, where  $y_k = C_k (a_k^2 + b_k^2) / D_k$ ,

$$h_{step}(t) = \sum_{k=1}^{n} \frac{D_k}{a_k^2 + b_k^2} \left( 1 - e^{a_k t} \left( \cos(b_k t) - \frac{a_k + \gamma_k}{b_k} \sin(b_k t) \right) \right)$$

• Time when peak overshoot occurs in each section

$$t_{peak}^{k} = -\frac{1}{b_{k}} \left( \operatorname{atan} \left( \frac{\gamma_{k} b_{k}}{a_{k}^{2} + \gamma_{k} a_{k} + b_{k}^{2}} \right) + \pi \right)$$

• For computing gradients only, approximate peak overshoot time as a constant times the average of second-order peak overshoot times

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#### **Distance Measures for Properties**

- Deviation in Magnitude Response
  - Euclidean distance over each passband, stopband, and transition band
- Deviation from Linear Phase Response
  - Measure deviation from linear phase over the passband

$$\sigma_{phase} = \int_{\omega_2}^{\omega_1} \left( \angle H(j\omega) - m_{lp}\omega \right)^2 d\omega$$

• Optimal slope of the phase,  $m_{lp}$ , is a function of the passband interval  $(\omega_1, \omega_2)$  as well as the pole and zero pairs

$$m_{lp} = \frac{\int_{\omega_2}^{\omega_1} \angle H(j\omega) \omega d\omega}{\int_{\omega_2}^{\omega_1} \omega^2 d\omega} = \frac{3}{2(\omega_2^3 - \omega_1^3)} \sum_{k=1}^n (f_k(\omega_1) - f_k(\omega_2))$$

• Calculated by computer algebra software

- Deviation for Peak Overshoot: (h<sub>step</sub>(t<sub>peak</sub>) 1)<sup>2</sup>
- Deviation for Quality Factors:  $Q_{eff}$  0.5

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**Optimizing for Linear Phase and Peak Overshoot** 

### **Table 1: Fourth-Order Lowpass Filter**

	Initial	Final
Pole Pair 1	- 8.4149 <u>+</u> <i>j</i> 20.3153	-7.7918 <u>+</u> <i>j</i> 22.8984
Pole Pair 2	-20.3153 <u>+</u> j 8.4149	$-19.5623 \pm j \ 0.6255$
Cost Function	1.17	0.000047
Peak Overshoot	0.16%	0.08%

- *initial value*: fourth-order lowpass Butterworth filter
- *final value*: a hybrid filter
- phase response in passband became *nearly linear*
- one second-order section more sensitive to perturbations
- quality factors: {0.541, 1.31} -> {0.500, 1.55}

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#### **Design Example (continued)**



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#### **Design Example (continued)**

#### **Trade-off Magnitude Response for Step and Phase Response**



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### **Algebraic Verification of Formulas**

- Formula for partial fractions decomposition
- Formula for step response
- Optimal slope for linear phase

### **Numerical Validation of Formulas**

• Magnitude and phase formulas

## Validation of Synthesized Code

- Plot Mathematica and MATLAB formulas
- SQP MATLAB *constr* routine checks symbolic gradients

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### **Automating the Solution**

- Enter objective measures, distance measures, and constraints in computer algebra environment
- Choose an optimization technique
- Transform optimization problem to fit the technique
- Synthesize transformed problem into software
- Export solution to a system-level design environment

### Advantages

- Abstract design specification to a higher level
- Avoid errors in performing algebra and calculus
- Avoid errors in converting equations to source code

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