EE345M Quiz 1	Spring 2008	Solution	Page 1				
Jonathan W. Valvanc	First Name	:	Last Name:				
February 15, 2008, 1	0:00 to 10:50am						
(30) Question 1. Cor	isider a file system t	hat uses a file tran	slation table (FTT)				
(10) Part a) There is	s no external fragme	entation. If there a	re n free blocks at any locations on the				
disk, they all can be data) Space used for	the directory and th	of size n-1 blocks	(1 block for the F11 and n-1 blocks for fragmentation				
(5) Part h) Assume a	file has n data bloc	ks It takes one blo	<i>agine mation.</i>				
Average = 1 r	nore read	KS. It takes one <i>bio</i>					
Maximum = 1	more read (<i>this is</i>	a fast method to re	ad when access is random)				
(5) Part c) Consider the linked allocation scheme described in Lab 25. On average it takes n/2 block reads to access the data							
				The worst cas	e is the byte is in the	e last block, requir	ing n block reads to access the data
(10) Part d) Write a C function that returns a byte from a file at a random location. The important							
part of this solution is	s to divide the addre	ss into two parts.					
unsigned char eFile_Read(unsigned char numFTT, unsigned short address){							
eDisk_ReadBlock	(FFTbuf,numFTT);						
<pre>msb = (unsigned char)address>>8; // top part of address lsb = (unsigned char)address&0x00FF; // bottom part of address blocknum = FTTbuf[msb]; // address translation eDisk ReadBlock(Databuf, blocknum);</pre>							
				return Databuf[lsb];		
				}			
(20) Question 2. Choose the largest block size so that the average internal fragmentation is less 5%.							
First, approximate by	ignoring the 16-by	te overnead					
Internal frage	IS Hall a DIOCK 0.5°	$^{\prime} 2$ $*2^{n}/100000 < 0.05$	$2^{n} < 10000$ $2^{n} - 8102$ n - 13				
Hard way		2/100000 < 0.05,	2 < 10000, 2 = 0172, 11–15				
Each block ca	in store a total of	$(2^{n}-16)$ bytes					
The number of blocks needed		m = 100000/(2	ⁿ -16) (round up to next integer)				
Wasted space	is	16*m + 0.5*(2)	ⁿ -16)				
Internal fragn	nentation	(16*(100000/(2	$(2^{n}-16)) + 0.5*(2^{n}-16))/100000 < 0.05$				
Check solution for n=13		m = 100000/(8)	192-16) = 13				
		((16*13) + 0.5)	*(8192-16))/100000 < 0.05				
		(208 + 0.5*(81))	76))/100000 < 0.05				
$2^{n} - 2102 m -$	12	(208 + 4088)/1	00000 = 0.043 < 0.05				
2 - 8192, II-	15 offs at $\frac{1}{4}$ fs and $\frac{1}{8}$	fo					
(10) Question 5. Cut		15					
		\land \land					
	Gain 🔪 /						
		\setminus /					
		\ /					
		\setminus /					
	I V	V					

500 1000 2000 Frequency

0

Jonathan W. Valvano

(15) Question 4. The largest component is at 60 Hz (thus, this probably due to EM field pick up), and there is a harmonic at 180 Hz. There are many steps to reduce EM field pick up

Shielded box, and shielded cables, PCB with ground plane

Use twisted pair or shorter cables (reduce area for EM field pickup) Switch the transducer to differential type and use an instrumentation amp with good CMRR Reduce the size of the EM source (ground or shield the source), move it farther away

60 Hz notch digital filter (can't use a HPF or LPF because you might remove signal)

(25) Question 5. There are two valid approaches to solving this problem.

Solution 1. Analog ground method. First, design the system assuming a $\pm V$ supply. In this first step, analog ground is the same as digital ground. The HPF cutoff is 1 Hz, $f_c = 1/(2\pi R_1 C_1)$. Notice, since the HPF feeds into the +terminal of the op amp, the HPF sees a very high input impedance looking into the amplifier stage. If C_1 is 10 µF, then $R_1 = 1/(2\pi 1 \mu F) = 160 \text{ k}\Omega$. Most any $R_1C_1=2\pi$ will be OK (we would like $1k\Omega \leq R_1 \leq 1M\Omega$). The range goes from 1V to 5V, so we need a gain of 5. In this circuit, the input (-0.5 to +0.5V) is converted to (-2.5 to +2.5V). Thus, we get $V_{out} = 5V_{in}-4V_g$, so $R_4/R_2=5$, $R_4/R_3=4$, choose feedback resistor, $R_4=100 \text{ k}\Omega$. (We would like $1k\Omega \leq R_2$, R_3 , $R_4 \leq 1M\Omega$). The LPF cutoff is 200 Hz. Design the Butterworth LPF using LPF.xls



To use rail-to-rail single supply op amps, we set analog ground to 2.50V using the REF03. This way, the transducer output, V_t , has a range of 2 to 3 volts relative to digital ground. The amplifier still has a gain of 5. $V_{out} = 5V_{in}$ (relative to analog ground). Relative to digital ground we get $V_{out} = 5V_{in} - 10$. $V_{ref} = 2.50V$

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Solution 2. Digital ground method. We will design the system rail-to-rail single supply op amps, but be very careful not to have negative voltages on any of the op amp terminals. To convert the input (-0.5 to +0.5V) to (0 to +5V), we need $V_{out} = 5V_{in} + 2.5$, which is $V_{out} = 5V_{in} + V_{ref}$. We will need a ground gain of -5, so the sum of all gains is equal to 1. We need to make $R_7/R_2=5$ (V_{in} gain), $R_7/R_3=5$ (V_g gain), $R_7/R_4=1$ (V_{ref} gain), choose feedback resistor, $R_5 = 100 \text{ k}\Omega$. (We would like 1k $\Omega \leq R_2$, R_3 , R_4 , $R_7 \leq 1M\Omega$). The HPF and LPF are similar to the first method.



The HPF in this case will see the R_2+R_4 impedance to ground, so the analysis of this circuit is hardware than the first one.

FYI here is an FFT with noise much less than the ADC resolution.

