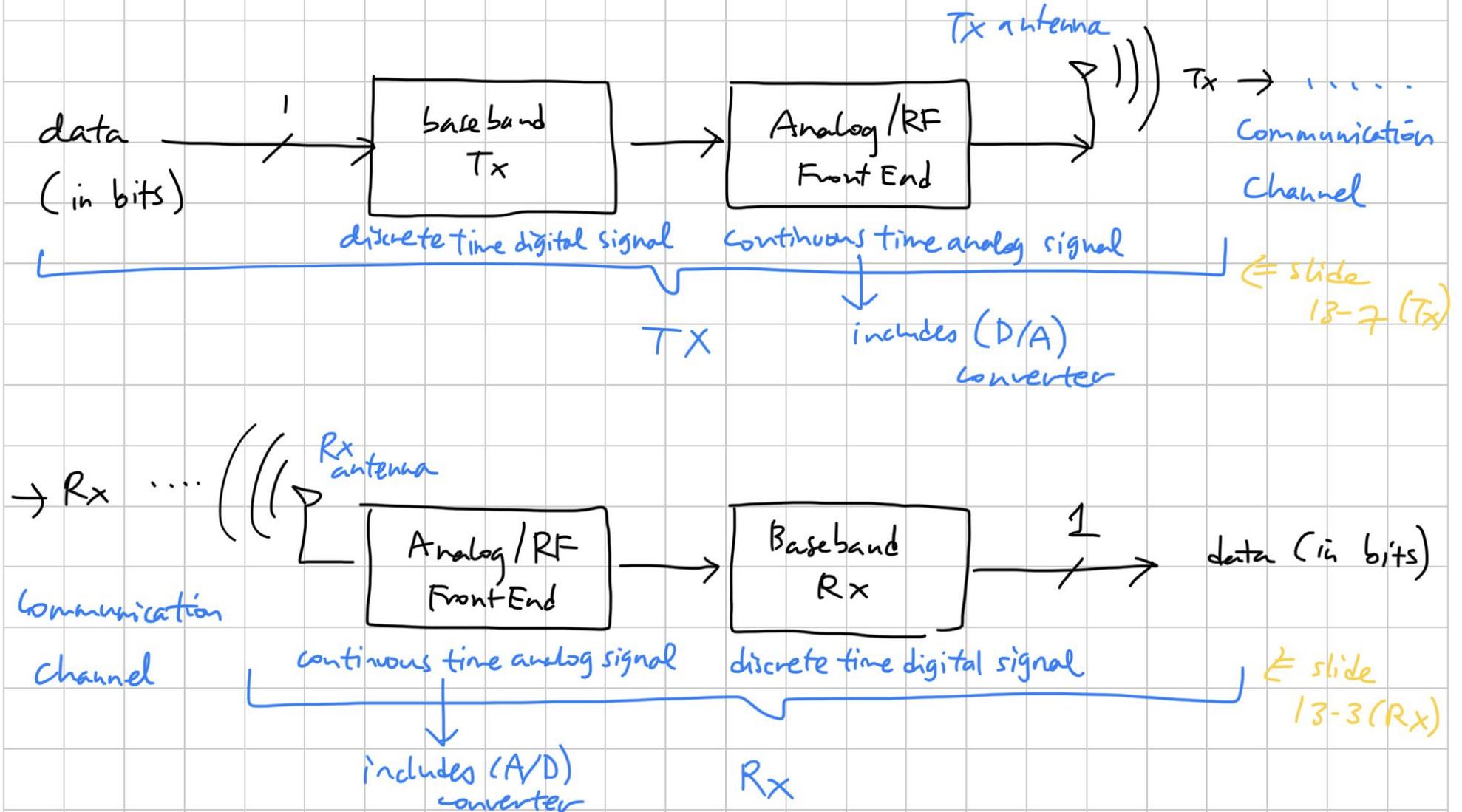


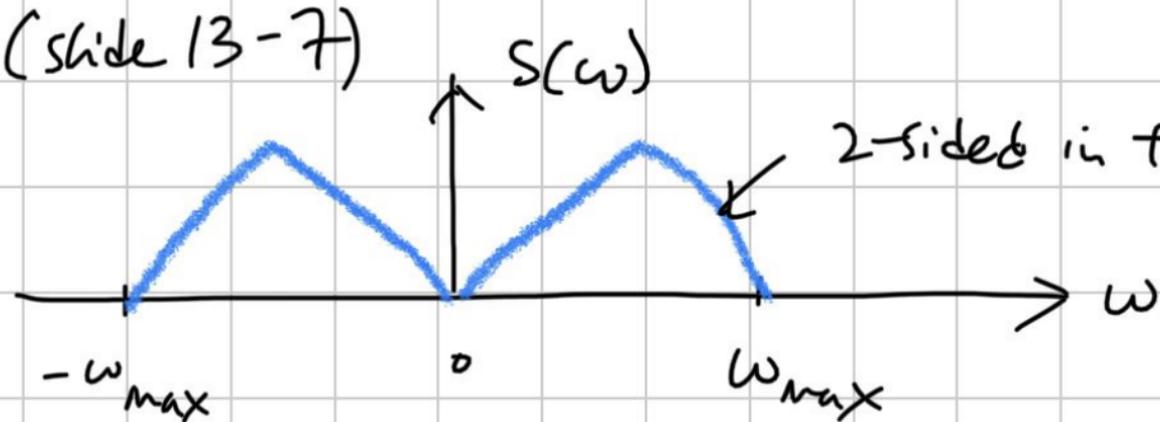
PHY (Physical) Layer (aka "Modem") Block Diagram



PAM Baseband (slide 13-7)



@ output of the pulse shaper



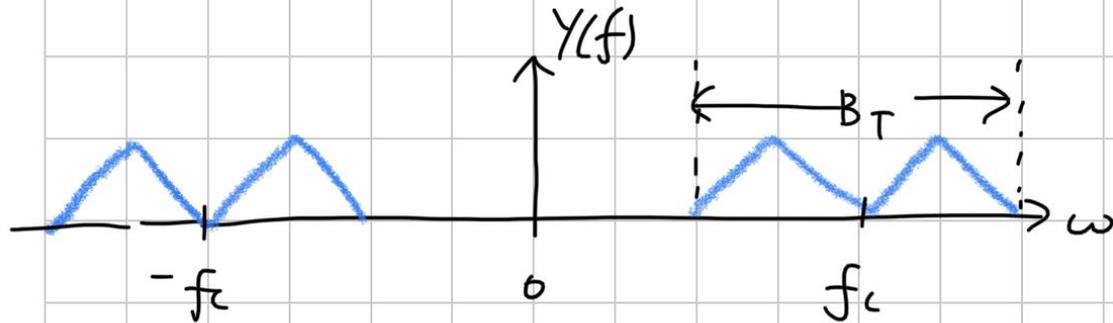
$$f_{\max} = \frac{1}{2} f_{\text{symbol}} (1 + \alpha) \quad \text{if pulse shaping = raised cosine pulse}$$

$\alpha \in [0, 1]$ = roll off factor

$$\omega_{\max} = 2\pi \frac{f_{\max}}{f_s}$$

then modulate to higher frequency.
we get PAM Band Pass as a result.

PAM Band Pass, $Y(f)$



$$B_T = 2 \left(\frac{1}{2} f_{\text{symp}} \right) (1 + \alpha) = f_{\text{symp}} (1 + \alpha)$$

★ you have limited
amount of Bandwidth to Tx over.
that is f_{symp} .

⇓
★ the other parameter we
have is the bit rate = $J f_{\text{symp}}$
 $J = \# \text{ bits / symbol you have.}$

let's go back to slide 13-3:

for a voltage of $V=d$, what is its power P ? $\Rightarrow P=V^2 = d^2$

ex: so if $V=3d$, $P=(3d)^2 = 9d^2$ (4PAM max power)

if $V=5d$, $P=(5d)^2 = 25d^2$ (6PAM max power)

\Rightarrow we see for each constellation of M-PAM,
it has a power budget. (set by regulations)

\Rightarrow thus, how to select "d" s.t. ① we satisfy power budget
② can distinguish between your
M symbols.

because you Tx symbols, say "d"

the communication channel adds noise (ideal scenario),

the Rx gets "d + n" (symbol + noise)

you need to select d small enough you satisfy max power budget
large enough to distinguish all M symbols
even in presence of noise

consider... you have max Tx Power

$$(M-1)d \text{ max symbol amplitude (V)}$$
$$(M-1)^2 d^2 \text{ max symbol power (W)}$$

→ for example, let max Tx Power = 16W

for 2PAM ($M=2$): $(M-1)^2 d^2 \leq 16W$

$$(2-1)^2 d^2 = d^2 \leq 16W$$
$$d \leq 4V$$

for 4PAM ($M=4$): $(M-1)^2 d^2 \leq 16W$

$$(4-1)^2 d^2 = 3^2 d^2 = 9d^2 \leq 16W$$
$$d \leq \frac{4}{3} V \approx 1.33V$$

then, the Rxer feeds back its Rxed SNR (signal-to-noise ratio)



to the Txer based on the Txed training signal

then the Txer can decide how many bits (J) to TX.

standards tells you what settings to have on the Txer side.

↳ i.e. LTE, WiFi, cellular comms.

(break @ 11:16 AM)

(return @ 11:22 AM)

Q: how to pick bits J ?

A: using Shannon Capacity Bound.

goal: maximize capacity (bits/second per Hz)

thus, $J \leq \lfloor \log_2 (1 + \text{SNR}) \rfloor$

$\lfloor \text{SNR in linear units (not dB scale)} \rfloor$

$\lfloor \cdot \rfloor = \text{floors to integer.}$

(ie. if $\log_2(1 + \text{SNR}) = 2.5$

then $J = \lfloor 2.5 \rfloor = 2$)

Polyphase Filter Bank vs Upsampling & Interpolation Filter
(Slide 13-16)

Direct Structure

↳ provides savings on roughly a factor of L in terms of computations & Memory size

walkthrough /
explanation of Polyphase Filter bank example compared to Direct Structure
(highly recommended to rewatch lecture @ 11:30 AM)

Polyphase Filter bank does the EXACT SAME result as the Direct Structure
but it removes all multiplications by 0

⇒ thus, the polyphase filters can operate at a lower frequency
(f_{symb})

but the Direct Structure must operate at the upsampled
frequency ($2f_{\text{symb}}$)

Polyphase filter bank gives us massive savings in terms of implementation / computational complexity, but does the exact same calculation as the Direct structure!

⇒ thus, we like to always use Polyphase Filters

how else could we improve speed?

⇒ parallel computation in hardware (ie. using FPGA)