

[10:30] Quadrature Amplitude Modulation (QAM) Receivers Part 2 (Review)

- A baseband signal in the transmitter experiences impairments as it passes through the transmitter analog/RF front end, channel, and receiver analog/RF front end
- A receiver would have subsystems to compensate impairments (Lecture Slide 16-3)
 - Fading: Automatic gain control
 - Additive noise: Matched filtering
 - Linear distortion: Channel equalizer
 - Carrier mismatch: Carrier recovery
 - Symbol timing mismatch: Symbol clock recovery (a.k.a. symbol timing recovery)

[10:40] Carrier Detection (Lecture Slide 16-9)

- Detect energy in the received signal
- Turn off subsystems when no transmission is occurring to save power
- Input: $x[m] = r^2[m] =$ instantaneous power
- Output: $p[m] = LPF\{x[m]\} =$ time-averaged power
- Low-complexity first-order IIR lowpass filter: $p[m] = c p[m - 1] + (1 - c) x[m]$
 - Since $p[m] \geq 0$ and $x[m] \geq 0$, $0 < c < 1$
 - Transfer function: $P(z) = c z^{-1}P(z) + (1 - c) X(z) \Rightarrow \frac{P(z)}{X(z)} = \frac{1-c}{1-cz^{-1}}$
 - Pole location c : larger $c \rightarrow$ more selective lowpass filter (more smoothing)
 - Choice of c can improve run-time efficiency
 - Example: $c = 3/4 \Rightarrow \frac{3}{4}p[m - 1] = \frac{1}{2}p[m - 1] + \frac{1}{4}p[m - 1]$
 - Since $p[m - 1] \geq 0$, multiplying $p[m-1]$ by 2^{-n} can be replaced by right shift by n if $p[m - 1]$ is in two's complement integer format

EE 445S Nov. 30, 2020

- QAM Receivers Part 2 (Lecture 16)
- Quantization (Lecture 8)

Carrier Detection
 $x[m] = r^2[m] \leftarrow$ instantaneous power
 $x[m] \rightarrow$ [LTI Filter] $\rightarrow p[m]$ average power
 IIR or FIR

IIR - LTI
 $p[m] = c p[m-1] + (1-c)x[m]$
 pole ≥ 0 ≥ 0

$P(z) = c z^{-1}P(z) + (1-c)X(z) \Rightarrow \frac{P(z)}{X(z)} = \frac{1-c}{1-cz^{-1}}$
 $0 < c < 1$

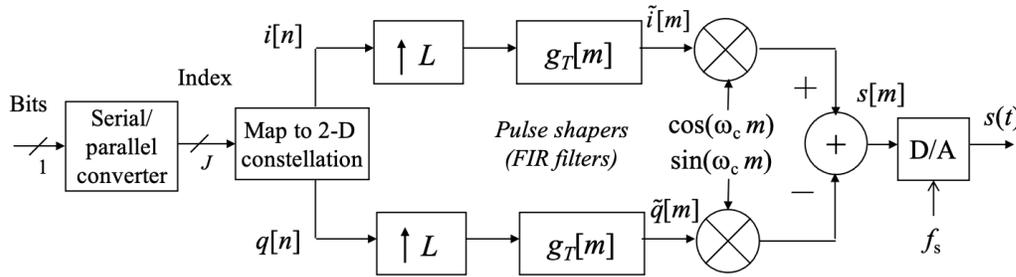
Let $c = \frac{3}{4}$ and $p[m-1] \geq 0$
 $\frac{3}{4}p[m-1] = \frac{1}{2}p[m-1] + \frac{1}{4}p[m-1]$
 ($p[m-1] \gg 1$)
 if $p[m-1]$ is two's complement integer

Lowpass filter
 $c \rightarrow$ 1 more Smoothing

Diagram of the z-plane showing a pole at $z=c$ on the real axis. The horizontal axis is labeled $Re\{z\}$ and the vertical axis is $Im\{z\}$. A circle is drawn around the origin, and the pole is marked with an 'x' at $z=c$.

[11:00] Symbol clock recovery (Lecture Slide 16-10)

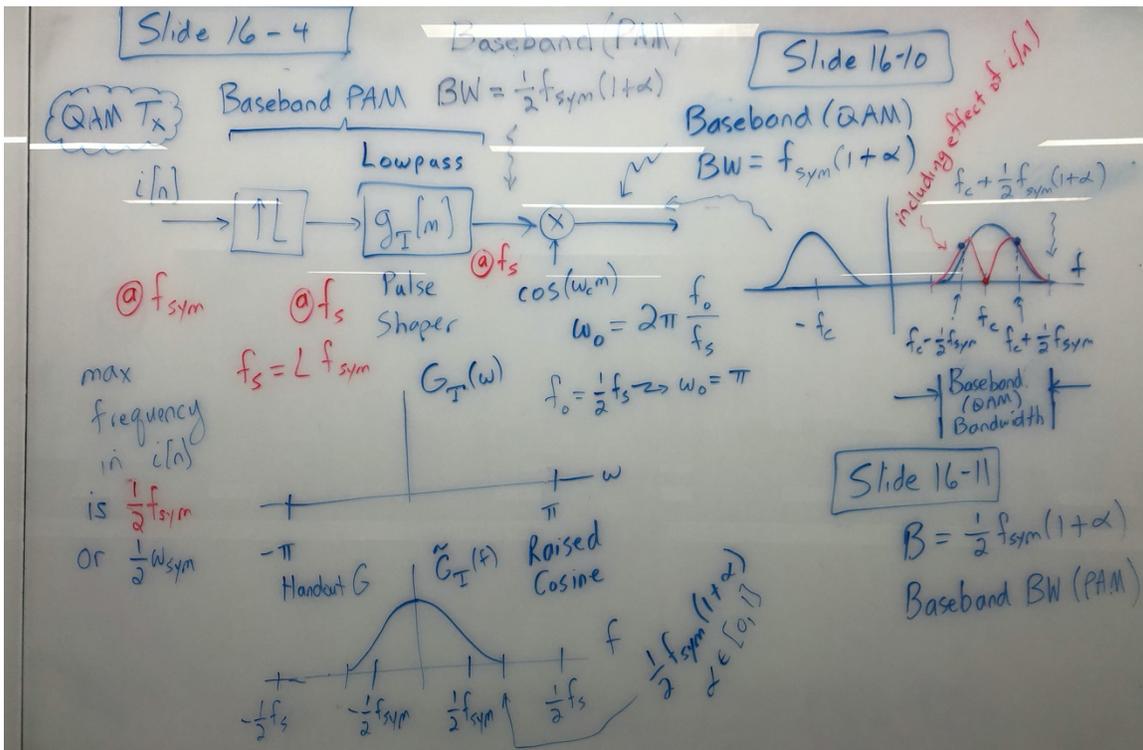
- QAM baseband transmitter consists of two upconverted Pulse Amplitude Modulation (PAM) transmitters that are 90 degrees out of phase (Lecture Slide 16-4)



- Recall structure of baseband PAM for in-phase part of baseband QAM transmitter:

$$i[n] \xrightarrow{\text{@ } f_{sym}} \uparrow L \xrightarrow{\text{@ } f_s} \underbrace{g_T[m]}_{\text{pulse shaper}} \xrightarrow{\text{@ } f_s} \tilde{i}[m]$$

- Lowpass pulse shaping filter should pass frequencies up to $f_{sym}/2$
- Raised cosine pulse shaping filter will expand bandwidth by factor $(1 + \alpha)$
- Baseband PAM bandwidth $B = \frac{1}{2} f_{sym} (1 + \alpha)$
- Upconversion by sinusoidal modulation doubles bandwidth to $f_{sym}(1 + \alpha)$
- $i[n]$ are symbol amplitudes with an average value of 0 (hence DC component is 0)
- One approach for symbol timing recovery (Handout M)
 - Use symmetry of received signal about f_c
 - Use two complex single-pole bandpass IIR filters with center frequencies $f_c - B$ and $f_c + B$ in parallel and combine outputs nonlinearly to estimate symbol timing offset



[11:25] Baseband QAM demodulation (Lecture Slides 16-11 and 16-12)

- Demodulation: multiply by sinusoid and lowpass filter
 - In phase: $\bar{i}[m] = LPF\{2 x[m] \cos(\omega_c m)\}$
 - Quadrature: $\bar{q}[m] = LPF\{-2 x[m] \sin(\omega_c m)\}$
- Lowpass filter serves multiple purposes
 - Matched filter
 - Anti-aliasing
 - Demodulation filter
- Downsample by L to return to symbol rate

