

# **Blind Estimation of FIR Channels in CDMA Systems with Aperiodic Spreading Sequences**

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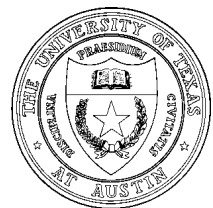
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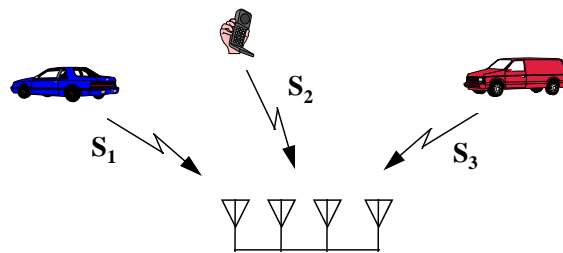
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# Motivations

- In multipath environment, CDMA receivers often suffer from interference.



- Blind signal estimation schemes cannot be used because they require periodic spreading sequences.
- RAKE receivers are often used, but they cannot fully exploit the rich structure of CDMA signals to minimize interference.

## Background: Channel Model

Multipath channel model between the  $i$ th user and the  $M$ -element antenna array at the base station

$$\mathbf{h}_i(t) = \begin{bmatrix} h_{1,i}(t) \\ \vdots \\ h_{M,i}(t) \end{bmatrix} = \sum_{l=1}^{L_i} \mathbf{a}_i(\theta_l) p(t - \tau_i(l))$$

- $p(t)$  is the pulse shaping function
- $\tau_i(l)$  is the delay
- $\mathbf{a}_i(\theta_l)$  is the array response vector of the  $l$ th multipath signal
- $L_i$  is the total number of paths of  $i$ th user

## Background: Data Model

The baseband signals from the antenna outputs of an asynchronous CDMA system with  $P$  users

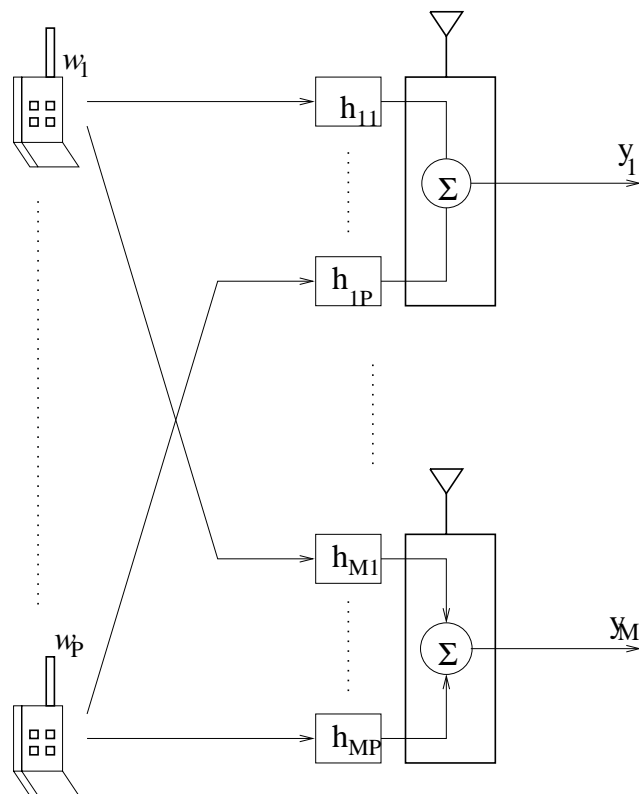
$$\mathbf{y}(t) = \sum_{i=1}^P \sum_{n=-\infty}^{\infty} w_i(n) \mathbf{h}_i(t - nT) + \mathbf{v}(t)$$

- $T$  is the chip period
- $\mathbf{v}(t)$  is the noise vector
- $w_i(k) = s_i(n)c_i(k - nL_c - k_i)$
- $n = \lfloor \frac{k - k_i}{L_c} \rfloor$ ;  $k_i$  ( $0 \leq k_i < L_c$ ) is the chip delay index assumed to be known.

### We know

- channels are FIR
- Source symbols are drawn from a finite alphabet
- Pseudo-noise (PN) spreading codes

# Channel Model in a P-User CDMA System with an M-element Antenna Array



# Blind Estimation Problem in CDMA Systems

- The blind estimation problem is to estimate FIR channel parameters without the use of training sequences.
- Many subspace-based algorithms have been successfully developed on multiuser FIR channel estimation in CDMA systems with periodic spreading sequences.
- These algorithms rely on the periodicity of spreading sequences in order to estimate the channel parameters.
- Furthermore, periodicity also simplify the use of multi-user detection techniques.
- **However** these algorithms are only applicable to CDMA systems with periodic spreading sequences.

## **Why Aperiodic Spreading Sequences?**

Many practical systems such as IS-95 use aperiodic spreading sequences to

- achieve uniform signal spectrum
- identify cell sites uniquely
- obtain other desirable properties.

A few algorithms have been developed on channel estimation to be used in such systems

Here, we propose a method which provides promising signal estimates using the inherent structure information of CDMA signals.

# Channel Estimation Methods

**Current Methods** in CDMA systems with aperiodic spreading sequences

- 1-D RAKE receivers
- 2-D RAKE receivers (with an antenna array)
- Principal Component (PC) Algorithm proposed by Liu and Zoltowski

## Our method

- Uses two different frameworks to capture the rich structure of CDMA signals



## First Framework at Chip Level

In the first framework, we construct the data matrix of the signal sampled at the chip rate

$$\mathbf{Y} = \mathbf{H}\mathbf{W} = \begin{bmatrix} \mathbf{h}_1 & \cdots & \mathbf{h}_P \end{bmatrix} \begin{bmatrix} \mathbf{w}_1(N) \\ \vdots \\ \mathbf{w}_P(N) \end{bmatrix}$$

- $\mathbf{h}_i = [\mathbf{h}_i(L - 1) \ \mathbf{h}_i(L - 2) \ \cdots \ \mathbf{h}_i(0)]$
- $\mathbf{w}_i$  is constructed as

$$\begin{bmatrix} w_i(1) & w_i(2) & \cdots & w_i(NL_c - L + 1) \\ w_i(2) & w_i(3) & \cdots & w_i(NL_c - L + 2) \\ \vdots & \vdots & \cdots & \vdots \\ w_i(L) & w_i(L + 1) & \cdots & w_i(NL_c) \end{bmatrix}$$

Solution for the above equation

$$\mathbf{H} = \mathbf{Y}\mathbf{W}^\dagger$$

Most of  $\mathbf{W}$  is known due to the known PN spreading sequence.

## Second Framework at Symbol level

In the second framework, we stack the spatial data samples so that the data matrix

$$\mathcal{Y} = \begin{bmatrix} \mathbf{y}^1 & \mathbf{y}^2 & \cdots & \mathbf{y}^M \end{bmatrix}^T$$

can be represented as

$$\mathcal{Y} = \underbrace{\begin{bmatrix} \mathcal{G}_1 & \mathcal{G}_2 & \cdots & \mathcal{G}_P \end{bmatrix}}_{\mathcal{G}} \underbrace{\begin{bmatrix} \mathbf{s}_1^T \\ \mathbf{s}_2^T \\ \vdots \\ \mathbf{s}_P^T \end{bmatrix}}_{\mathbf{S}}$$

- $\mathbf{s}_i = [s_i(1) \cdots s_i(N)]$ .

- $\mathcal{G}_i = \underbrace{\begin{bmatrix} \mathcal{C}_i & \mathbf{0} \\ & \ddots \\ \mathbf{0} & \mathcal{C}_i \end{bmatrix}}_{M \text{ blocks}} \begin{bmatrix} \mathcal{H}_i^1 \\ \vdots \\ \mathcal{H}_i^M \end{bmatrix}$



## Definition of $C_i(n)$ and Solution for $S$

A complete block of  $C_i(n)$   $n = 0, \dots, N - 1$  can be written as

$$\underbrace{\left[ \begin{array}{ccc} 0 & \dots & 0 \\ \dots & \ddots & \dots \\ c_i(nL_c + 1) & \ddots & 0 \\ \vdots & \ddots & c_i(nL_c + 1) \\ c_i(nL_c + L_c) & \ddots & \vdots \\ 0 & \ddots & c_i(nL_c + L_c) \\ \vdots & \dots & 0 \\ 0 & \dots & 0 \end{array} \right]}_{L \text{ columns}} \left. \vphantom{\left[ \begin{array}{ccc} 0 & \dots & 0 \\ \dots & \ddots & \dots \\ c_i(nL_c + 1) & \ddots & 0 \\ \vdots & \ddots & c_i(nL_c + 1) \\ c_i(nL_c + L_c) & \ddots & \vdots \\ 0 & \ddots & c_i(nL_c + L_c) \\ \vdots & \dots & 0 \\ 0 & \dots & 0 \end{array} \right]} \right\} 2L_c$$

If we use the equation  $\mathcal{Y} = \mathcal{G}S$  to solve for  $S$ , then we get

$$\boxed{S = \mathcal{G}^\dagger \mathcal{Y}}$$

## Definition Of The Algorithm

- $\mathbf{H} = \mathbf{Y}\mathbf{W}^\dagger$  and  $\mathbf{S} = \mathcal{G}^\dagger\mathcal{Y}$  allow us to use both frameworks to exploit the discrete-alphabet property of CDMA signals and knowledge of spreading codes.
- We adopt an Iterative Least Squares with Projection (ILSP) algorithm originally developed by Talwar, Viberg, and Paulraj for TDMA systems.
- We update  $\mathbf{S}$  iteratively, which updates  $\mathbf{W}$ , and  $\mathbf{H}$  under the constraint that the information symbols  $\mathbf{S}$  are from finite alphabets.
- We continue to iterate until  $\mathbf{S}$  or  $\mathbf{H}$  converge.

## Algorithm Outline

1. Randomly choose  $\mathbf{S}_0$  and set  $l = 0$

2.  $l := l + 1$

(a)  $\left[ \mathbf{h}_{1,l} \quad \cdots \quad \mathbf{h}_{P,l} \right] = \mathbf{Y} \mathbf{W}_l^\dagger$  where

$$\mathbf{W}_l = \begin{bmatrix} \mathbf{w}_{l,1}(NL_c) \\ \vdots \\ \mathbf{w}_{l,P}(NL_c) \end{bmatrix}$$

(b) Construct  $\mathcal{G}_l$  with the estimated channel parameters and PN sequences.

(c)  $\mathbf{S}_{l+1}$  is estimated through

$$\mathbf{S}_{l+1} = \mathcal{G}_l^\dagger \mathcal{Y}.$$

(d) Project  $[s_{l,i}(k)]$  to closest discrete values.

3. Continue until  $\mathbf{S}_{l+1} - \mathbf{S}_l = \mathbf{0}$ .

## Simulation Results

We simulate a single receiver CDMA system with  $L_c = 16$ ,  $P = 8$  and a multi-receiver system with  $L_c$ ,  $M = 2$ ,  $P = 13$  and  $\text{SNR} = 15$  dB.

- Principal Component Algorithm
- Proposed Method

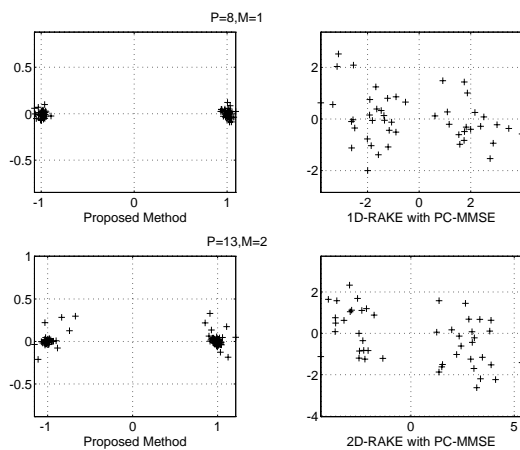


Figure 1: Signal constellations: 1-D RAKE, 2-D RAKE and our iterative method for the 1-D ( $M=1$ ) and 2-D ( $M=2$ ) cases.

## Simulation Results

Compare the mean square errors of our channel estimation and teh principal component (PC) algorithm.

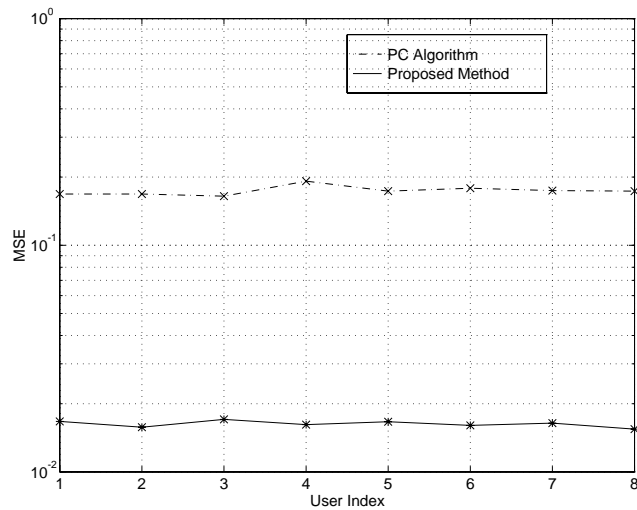


Figure 2: Comparison of two methods at SNR=15 dB

The proposed method offers better channel estimation.



# Simulation Results

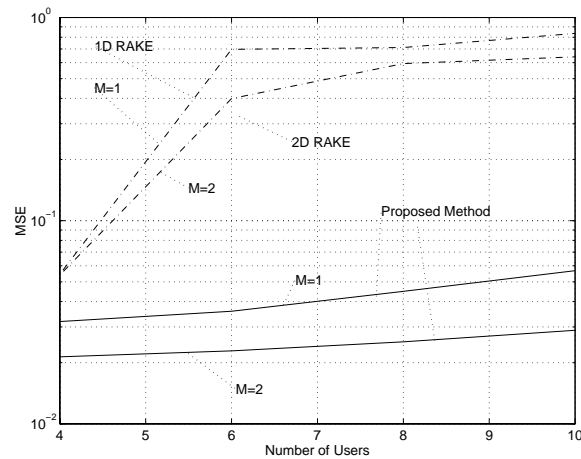


Figure 3: MSE vs. number of users using different receivers: 1-D RAKE, 2-D RAKE and our iterative method for the 1-D (M=1) and 2-D (M=2) cases.

# of Users	M=1	M=2
4	3.06	2.86
6	3.83	3.19
8	4.44	3.45
10	-	3.89
12	-	4.23
14	-	4.56
16	-	5.11

Table 1: Average Number of Iterations vs. Number of Antennas and Users

## Conclusions

- We present a new approach for blind estimation of FIR channels in CDMA systems with aperiodic spreading sequences.
- We derive two frameworks to exploit the structure information of CDMA signals efficiently,
- We develop an iterative technique based on iterative least-squares and projection.
- We perform computer simulations to demonstrate the effectiveness of the proposed scheme over existing methods.
- Our future directions include proving a necessary and sufficient condition for identifiability.